

LME49710 High-Performance, High-Fidelity Audio Operational Amplifier

Check for Samples: [LME49710](#)

FEATURES

- Easily Drives 600Ω Loads
- Optimized for Superior Audio Signal Fidelity
- Output Short Circuit Protection
- PSRR and CMRR Exceed 120dB (Typ)
- SOIC, PDIP, and TO-99 Packages

APPLICATIONS

- Ultra High-Quality Audio Amplification
- High-Fidelity Preamplifiers
- High-Fidelity Multimedia
- State-of-the-Art Phono Pre Amps
- High-Performance Professional Audio
- High-Fidelity Equalization and Crossover Networks
- High-Performance Line Drivers
- High-Performance Line Receivers
- High-Fidelity Active Filters

KEY SPECIFICATIONS

- Power Supply Voltage Range: $\pm 2.5V$ to $\pm 17V$
- THD+N ($A_V = 1$, $V_{OUT} = 3V_{RMS}$, $f_{IN} = 1kHz$)
 - $R_L = 2k\Omega$: 0.00003% (typ)
 - $R_L = 600\Omega$: 0.00003% (typ)
- Input Noise Density: $2.5nV/\sqrt{Hz}$ (typ)
- Slew Rate: $\pm 20V/\mu s$ (typ)
- Gain Bandwidth Product: 55MHz (typ)
- Open Loop Gain ($R_L = 600\Omega$): 140dB (typ)
- Input Bias Current: 7nA (typ)
- Input Offset Voltage: 0.05mV (typ)
- DC Gain Linearity Error: 0.000009%

DESCRIPTION

The LME49710 is part of the ultra-low distortion, low-noise, high-slew-rate operational amplifier series optimized and fully specified for high-performance, high-fidelity applications. Combining advanced leading-edge process technology with state-of-the-art circuit design, the LME49710 audio operational amplifiers deliver superior audio signal amplification for outstanding audio performance. The LME49710 combines extremely low-voltage noise density ($2.5nV/\sqrt{Hz}$) with vanishingly low THD+N (0.00003%) to easily satisfy the most demanding audio applications. To ensure that the most challenging loads are driven without compromise, the LME49710 has a high slew rate of $\pm 20V/\mu s$ and an output current capability of $\pm 26mA$. Further, dynamic range is maximized by an output stage that drives $2k\Omega$ loads to within 1V of either power supply voltage and to within 1.4V when driving 600Ω loads.

The LME49710's outstanding CMRR (120dB), PSRR (120dB), and V_{OS} (0.05mV) give the amplifier excellent operational amplifier DC performance.

The LME49710 has a wide supply range of $\pm 2.5V$ to $\pm 17V$. Over this supply range the LME49710's input circuitry maintains excellent common-mode and power supply rejection, as well as maintaining its low-input bias current. The LME49710 is unity gain stable. The Audio Operational Amplifier achieves outstanding AC performance while driving complex loads with values as high as 100pF.

The LME49710 is available in an 8-lead narrow body SOIC, an 8-lead PDIP, and an 8-lead TO-99. Demonstration boards are available for each package.



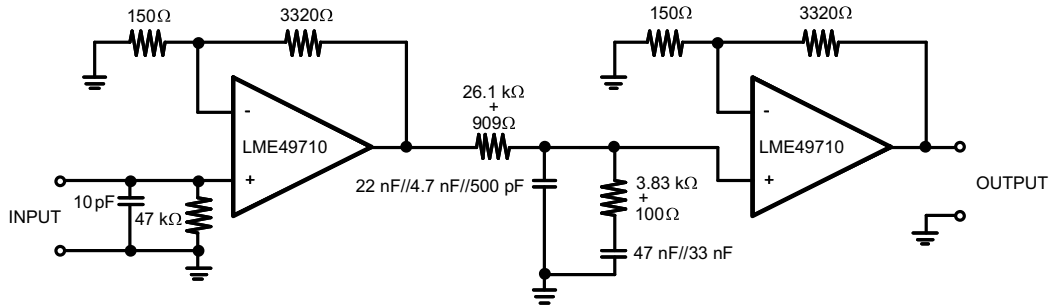
Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

All trademarks are the property of their respective owners.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of the Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

Copyright © 2006–2013, Texas Instruments Incorporated

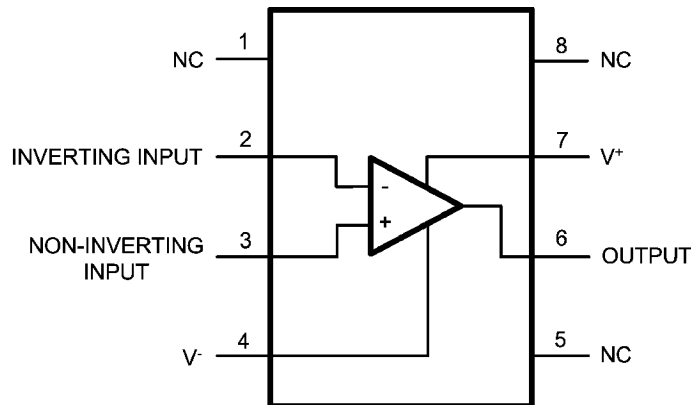
TYPICAL APPLICATION



Note: 1% metal film resistors, 5% polypropylene capacitors

Figure 1. Passively Equalized RIAA Phono Preamp

CONNECTION DIAGRAMS



**Figure 2. 8-Lead SOIC (D Package)
8-Lead PDIP (P Package)**

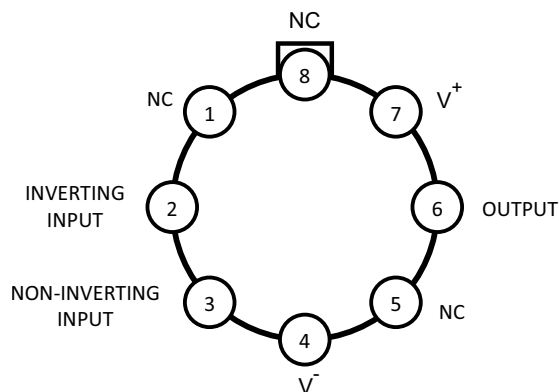


Figure 3. 8-Lead TO-99 (LMC Package)



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾⁽²⁾⁽³⁾

Power Supply Voltage ($V_S = V^+ - V^-$)		36V
Storage Temperature		-65°C to 150°C
Input Voltage		(V-) - 0.7V to (V+) + 0.7V
Output Short Circuit ⁽⁴⁾		Continuous
Power Dissipation		Internally Limited
ESD Susceptibility ⁽⁵⁾		2000V
ESD Susceptibility ⁽⁶⁾		200V
Junction Temperature		150°C
Thermal Resistance	θ_{JA} (D)	145°C/W
	θ_{JA} (P)	102°C/W
	θ_{JA} (LMC)	150°C/W
	θ_{JC} (LMC)	35°C/W
Temperature Range ($T_{MIN} \leq T_A \leq T_{MAX}$)		-40°C $\leq T_A \leq$ 85°C
Supply Voltage Range		$\pm 2.5V \leq V_S \leq \pm 17V$

- (1) *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur.
- (2) *Operating Ratings* indicate conditions for which the device is functional, but do not ensure specific performance limits. For ensured specifications and test conditions, see the Electrical Characteristics. The ensured specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.
- (3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (4) Amplifier output connected to GND, any number of amplifiers within a package.
- (5) Human body model, 100pF discharged through a 1.5k Ω resistor.
- (6) Machine Model ESD test is covered by specification EIAJ IC-121-1981. A 200pF cap is charged to the specified voltage and then discharged directly into the IC with no external series resistor (resistance of discharge path must be under 50 Ω).

ELECTRICAL CHARACTERISTICS⁽¹⁾⁽²⁾

The following specifications apply for $V_S = \pm 15V$, $R_L = 2k\Omega$, $f_{IN} = 1kHz$, and $T_A = 25^\circ C$, unless otherwise specified.

Symbol	Parameter	Conditions	LME49710		Units (Limits)
			Typical ⁽³⁾	Limit ⁽⁴⁾⁽⁵⁾	
THD+N	Total Harmonic Distortion + Noise	$A_V = 1$, $V_{OUT} = 3V_{RMS}$ $R_L = 2k\Omega$ $R_L = 600\Omega$	0.00003 0.00003	0.00009	% (max) % (max)
IMD	Intermodulation Distortion	$A_V = 1$, $V_{OUT} = 3V_{RMS}$ Two-tone, 60Hz & 7kHz 4:1	0.00005		% (max)
GBWP	Gain Bandwidth Product		55	45	MHz (min)
SR	Slew Rate		± 20	± 15	V/ μs (min)
FPBW	Full Power Bandwidth	$V_{OUT} = 1V_{P.P.}$, -3dB referenced to output magnitude at $f = 1kHz$	10		MHz
t_s	Settling time	$A_V = 1$, 10V step, $C_L = 100pF$ 0.1% error range	1.2		μs
e_n	Equivalent Input Noise Voltage	$f_{BW} = 20Hz$ to 20kHz	0.34	0.65	μV_{RMS}
	Equivalent Input Noise Density	$f = 1kHz$ $f = 10Hz$	2.5 6.4	4.7	nV/\sqrt{Hz} nV/\sqrt{Hz}
i_n	Current Noise Density	$f = 1kHz$	1.6		pA/\sqrt{Hz}
		$f = 10Hz$	3.1		pA/\sqrt{Hz}

- (1) *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur.
- (2) *Operating Ratings* indicate conditions for which the device is functional, but do not ensure specific performance limits. For ensured specifications and test conditions, see the Electrical Characteristics. The ensured specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.
- (3) Typical specifications are specified at +25°C and represent the most likely parametric norm.
- (4) Tested limits are specified to AOQL (Average Outgoing Quality Level).
- (5) Datasheet min/max specification limits are ensured by design, test, or statistical analysis.

ELECTRICAL CHARACTERISTICS⁽¹⁾⁽²⁾ (continued)

The following specifications apply for $V_S = \pm 15V$, $R_L = 2k\Omega$, $f_{IN} = 1kHz$, and $T_A = 25^\circ C$, unless otherwise specified.

Symbol	Parameter	Conditions	LME49710		Units (Limits)
			Typical ⁽³⁾	Limit ⁽⁴⁾⁽⁵⁾	
V_{OS}	Offset Voltage		± 0.05	± 0.7	mV (max)
$\Delta V_{OS}/\Delta Temp$	Average Input Offset Voltage Drift vs Temperature	$40^\circ C \leq T_A \leq 85^\circ C$	0.2		$\mu V/^\circ C$
PSRR	Average Input Offset Voltage Shift vs Power Supply Voltage	$\Delta V_S = 20V^{(6)}$	125	110	dB (min)
I_B	Input Bias Current	$V_{CM} = 0V$	7	72	nA (max)
$\Delta I_{OS}/\Delta Temp$	Input Bias Current Drift vs Temperature	$-40^\circ C \leq T_A \leq 85^\circ C$	0.1		nA/°C
I_{OS}	Input Offset Current	$V_{CM} = 0V$	5	65	nA (max)
V_{IN-CM}	Common-Mode Input Voltage Range		+14.1 -13.9	(V+) - 2.0 (V-) + 2.0	V (min) V (min)
CMRR	Common-Mode Rejection	$-10V < V_{CM} < 10V$	120	110	dB (min)
Z_{IN}	Differential Input Impedance		30		k Ω
	Common Mode Input Impedance	$-10V < V_{CM} < 10V$	1000		M Ω
A_{VOL}	Open Loop Voltage Gain	$-10V < V_{OUT} < 10V$, $R_L = 600\Omega$	140		dB
		$-10V < V_{OUT} < 10V$, $R_L = 2k\Omega$	140	125	dB
		$-10V < V_{OUT} < 10V$, $R_L = 10k\Omega$	140		dB
V_{OUTMAX}	Maximum Output Voltage Swing	$R_L = 600\Omega$	± 13.6	± 12.5	V
		$R_L = 2k\Omega$	± 14.0		V
		$R_L = 10k\Omega$	± 14.1		V
I_{OUT}	Output Current	$R_L = 600\Omega$, $V_S = \pm 17V$	± 26	± 23	mA (min)
I_{OUT-CC}	Short Circuit Current		+53 -42		mA mA
R_{OUT}	Output Impedance	$f_{IN} = 10kHz$ Closed-Loop	0.01		Ω
		Open-Loop	13		Ω
C_{LOAD}	Capacitive Load Drive Overshoot	100pF	16		%
I_S	Quiescent Current	$I_{OUT} = 0mA$	4.8	5.5	mA (max)

(6) PSRR is measured as follows: V_{OS} is measured at two supply voltages, $\pm 5V$ and $\pm 15V$. $PSRR = |20\log(\Delta V_{OS}/\Delta V_S)|$.

TYPICAL PERFORMANCE CHARACTERISTICS

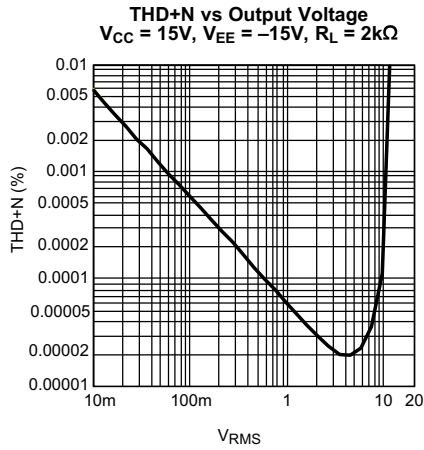


Figure 4.

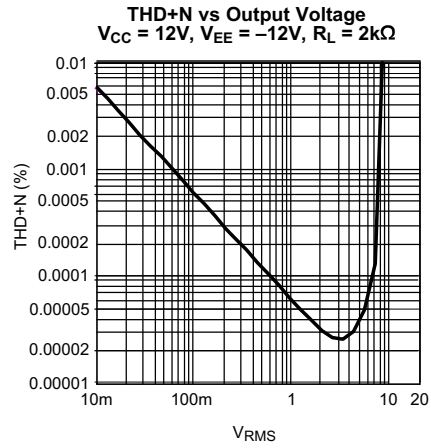


Figure 5.

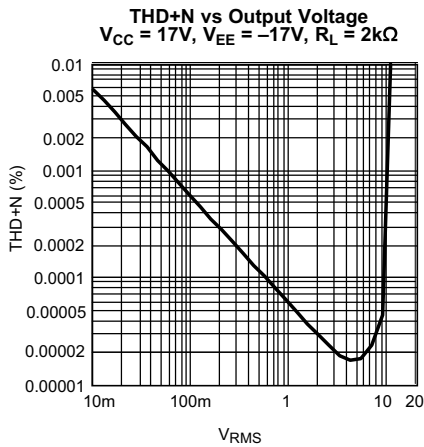


Figure 6.

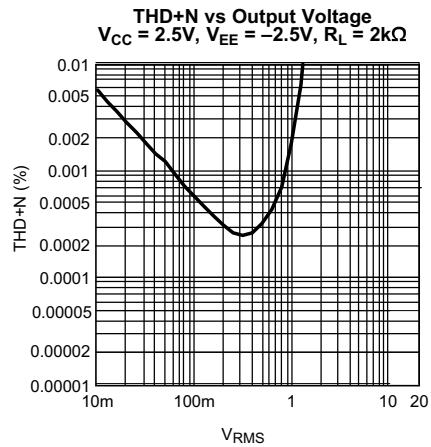


Figure 7.

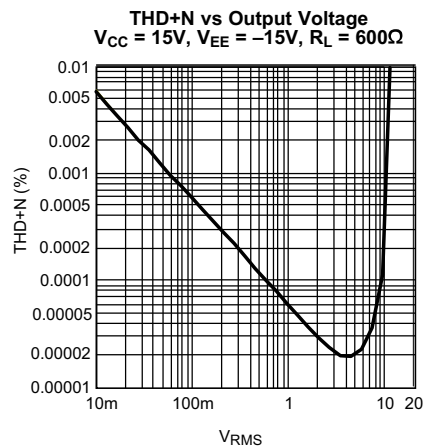


Figure 8.

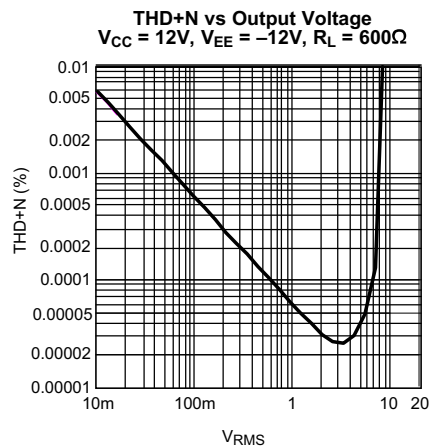


Figure 9.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

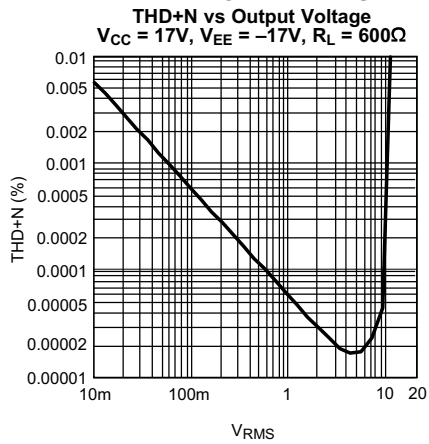


Figure 10.

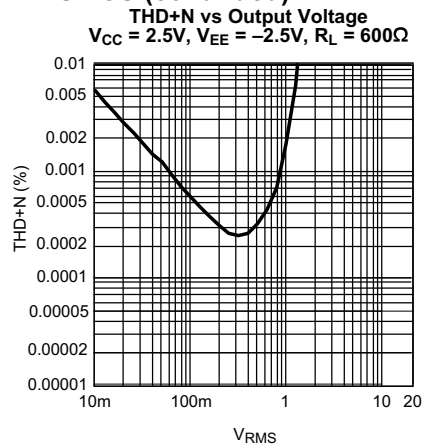


Figure 11.

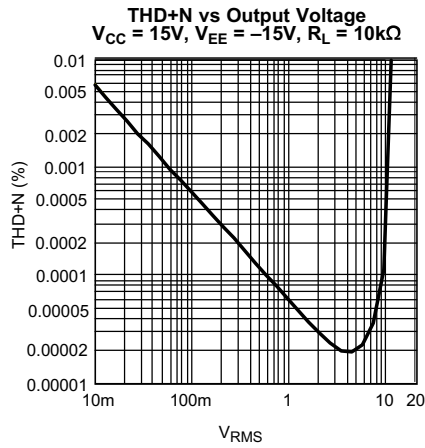


Figure 12.

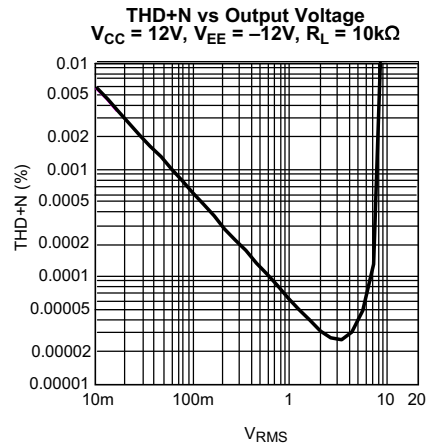


Figure 13.

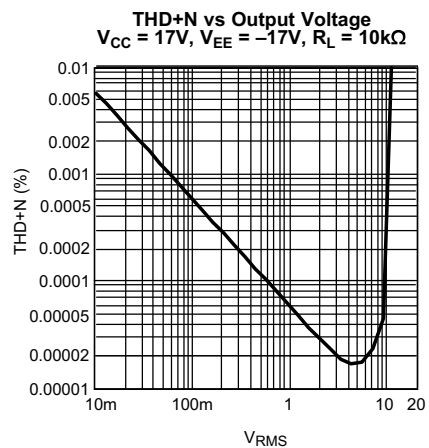


Figure 14.

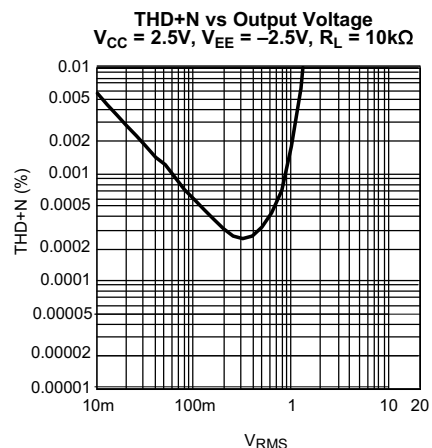


Figure 15.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

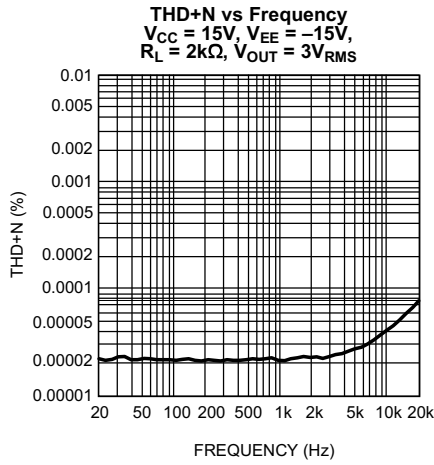


Figure 16.

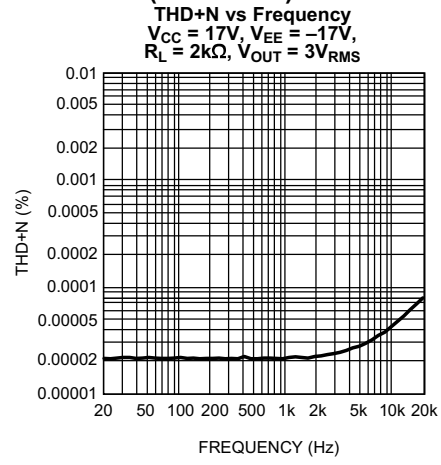


Figure 17.

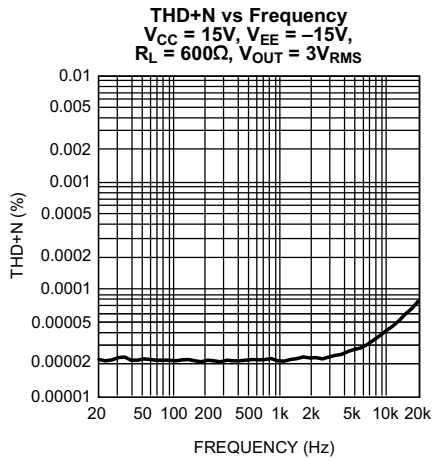


Figure 18.

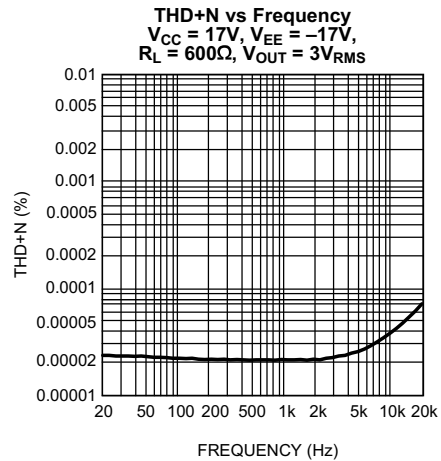


Figure 19.

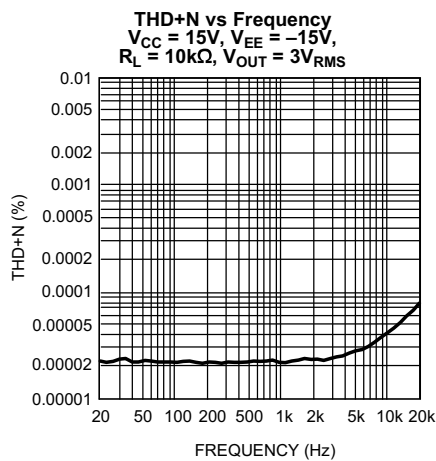


Figure 20.

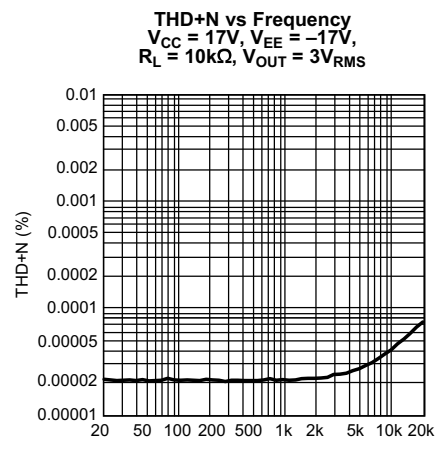


Figure 21.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

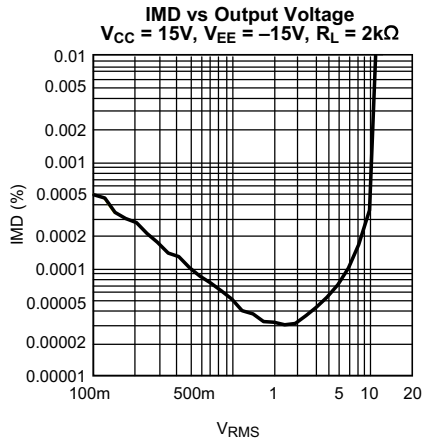


Figure 22.

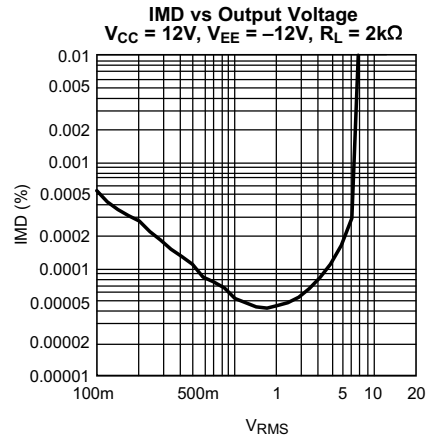


Figure 23.

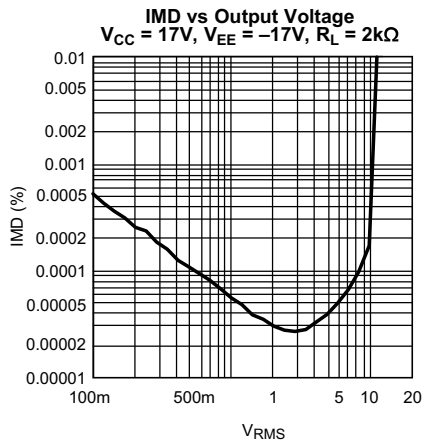


Figure 24.

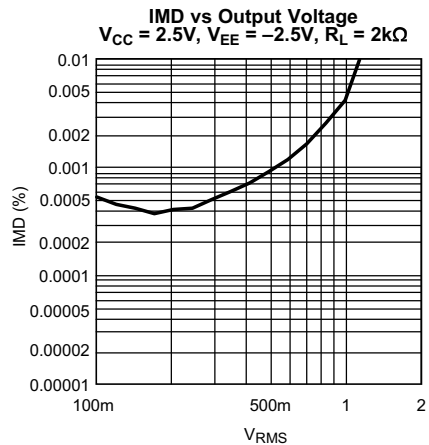


Figure 25.

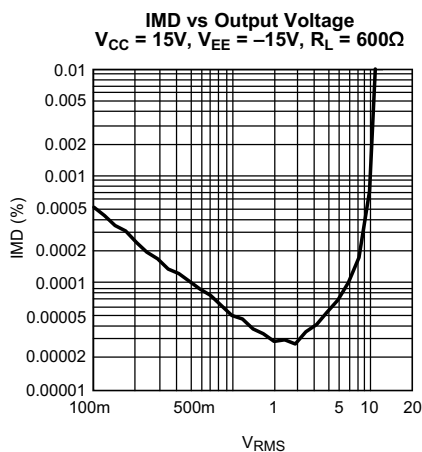


Figure 26.

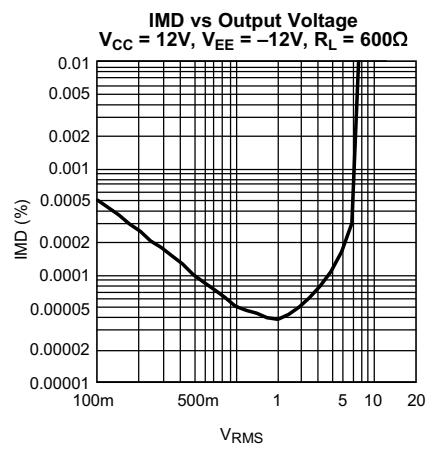


Figure 27.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

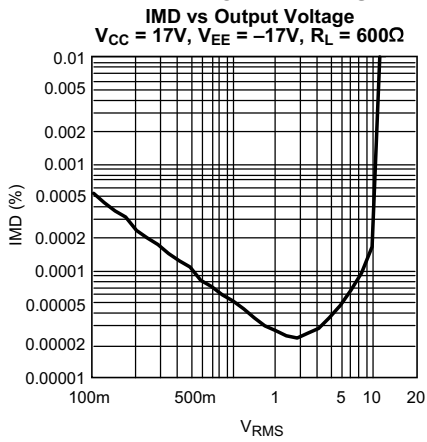


Figure 28.

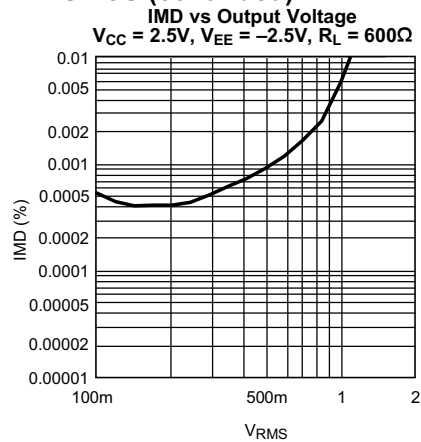


Figure 29.

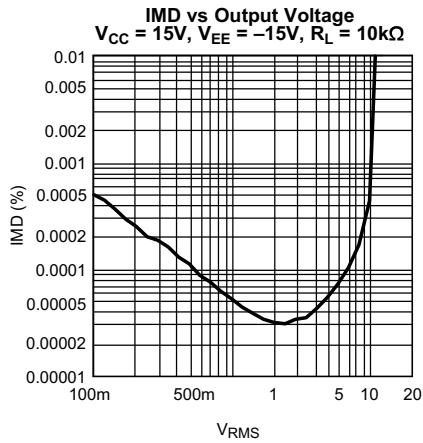


Figure 30.

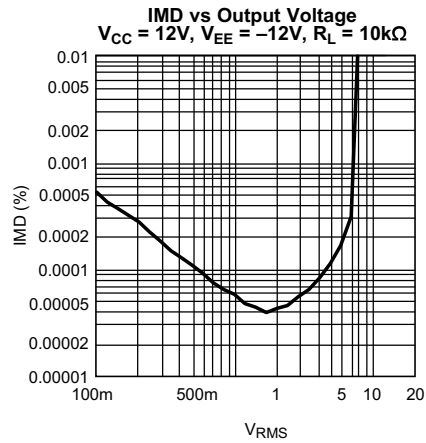


Figure 31.

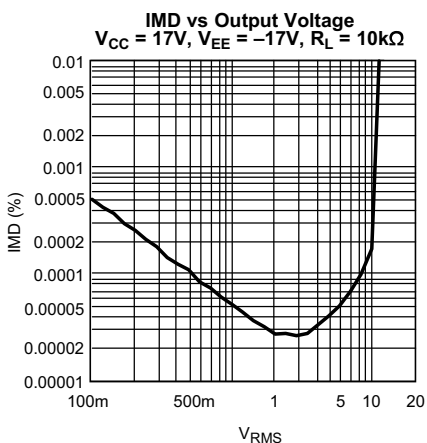


Figure 32.

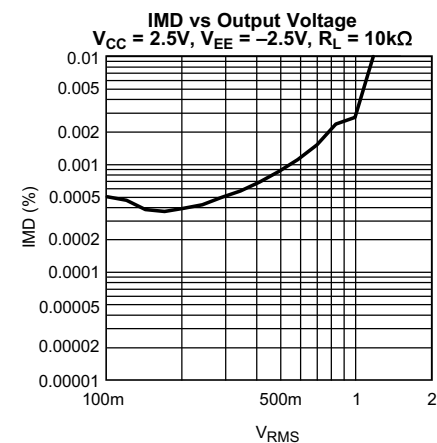


Figure 33.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

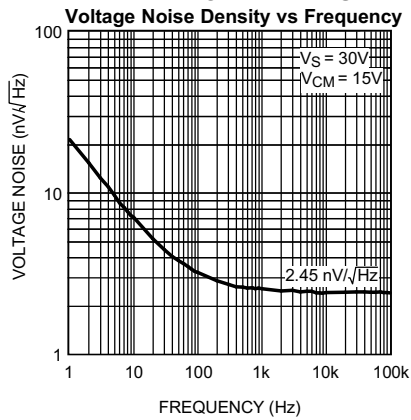


Figure 34.

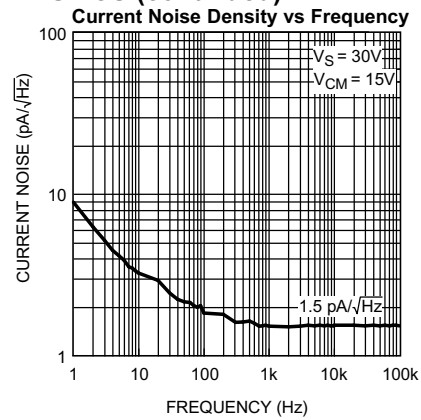


Figure 35.

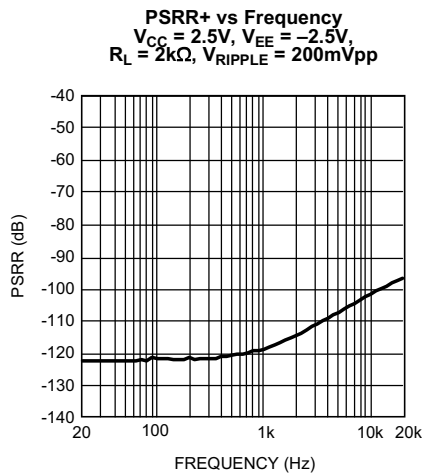


Figure 36.

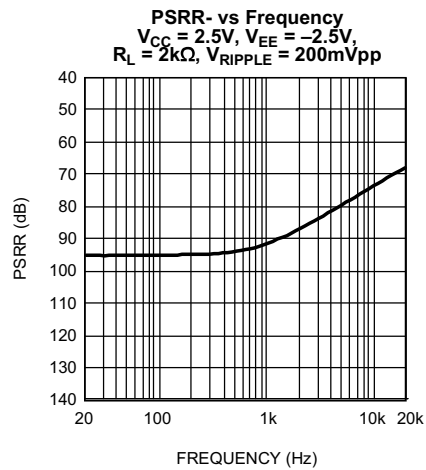


Figure 37.

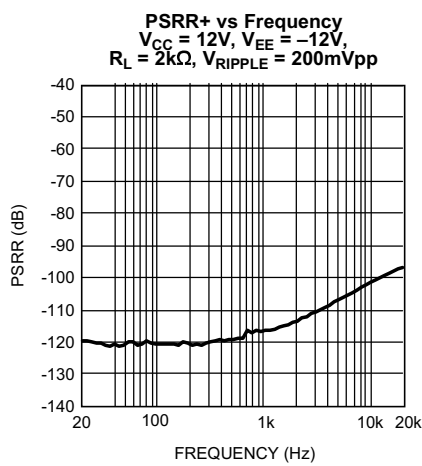


Figure 38.

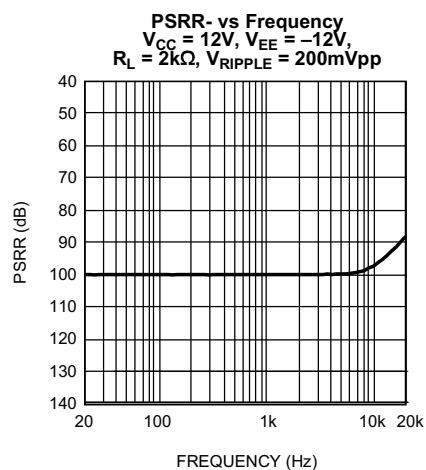


Figure 39.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

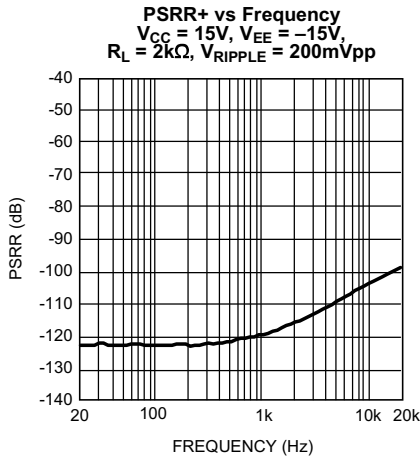


Figure 40.

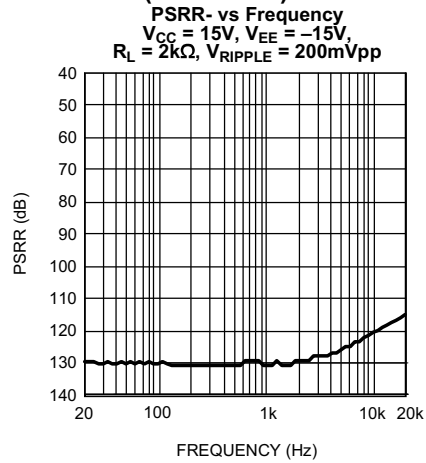


Figure 41.

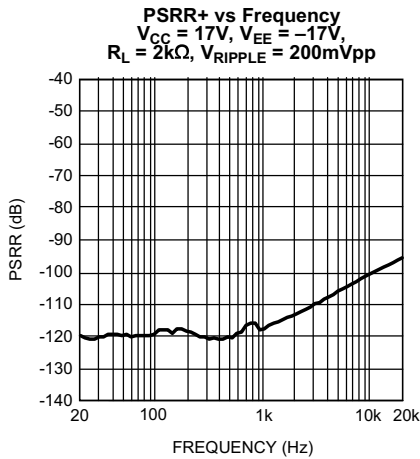


Figure 42.

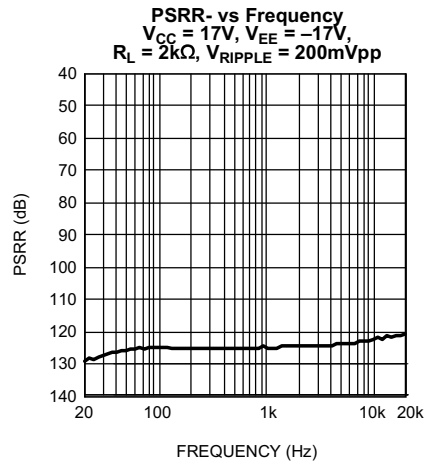


Figure 43.

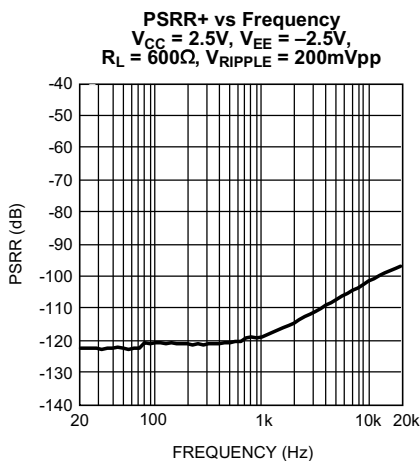


Figure 44.

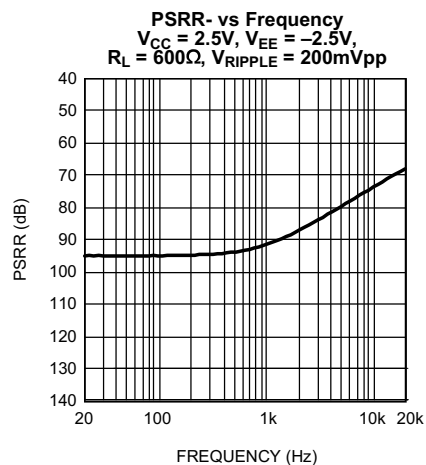


Figure 45.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

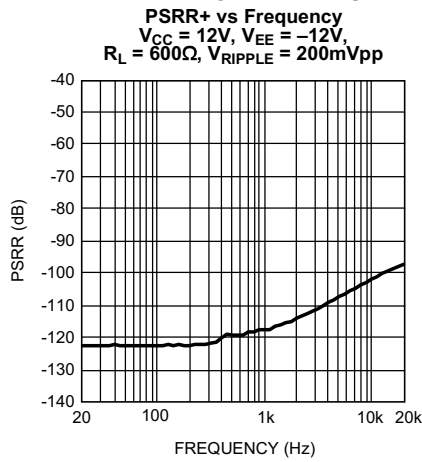


Figure 46.

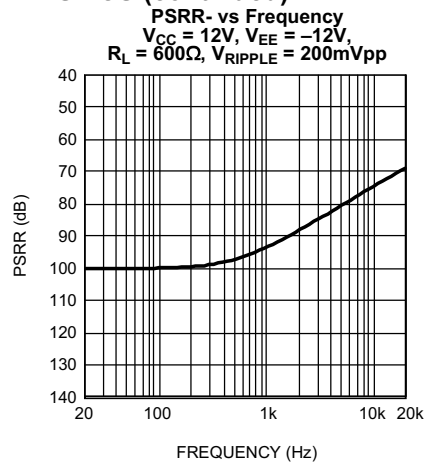


Figure 47.

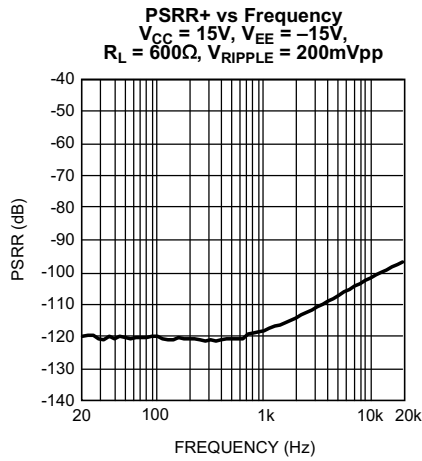


Figure 48.

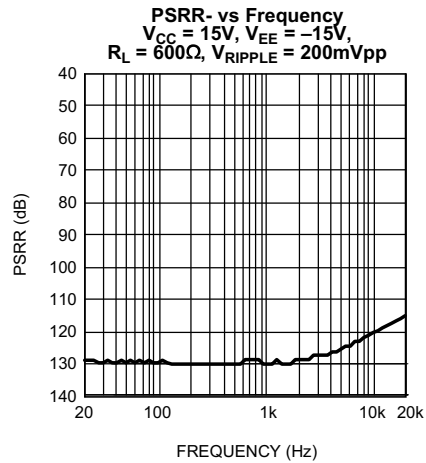


Figure 49.

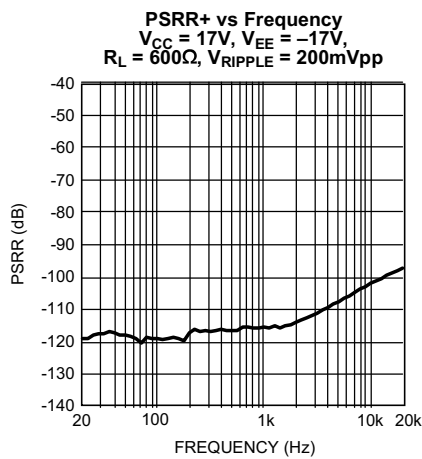


Figure 50.

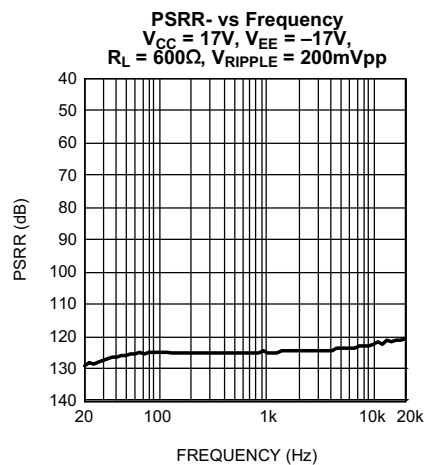


Figure 51.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

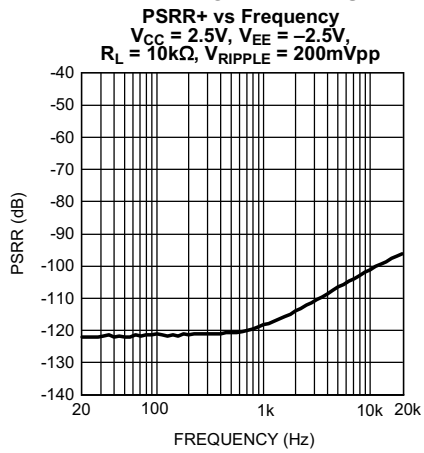


Figure 52.

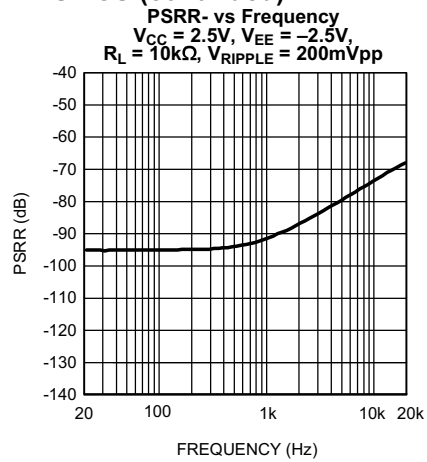


Figure 53.

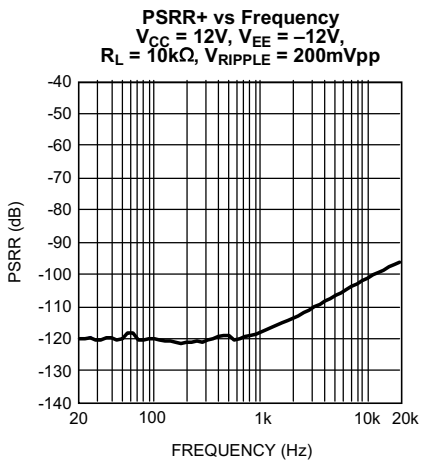


Figure 54.

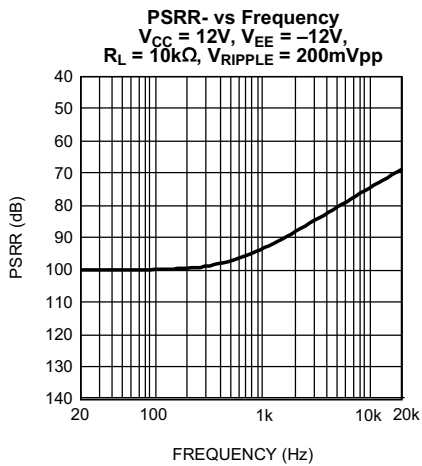


Figure 55.

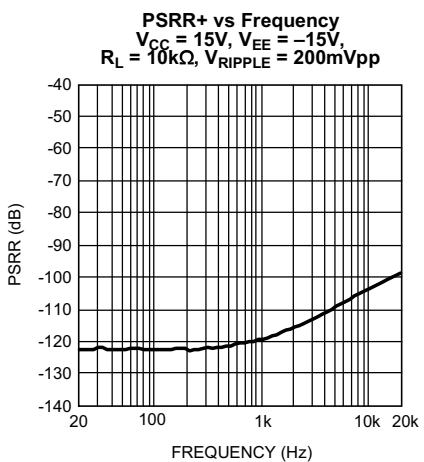


Figure 56.

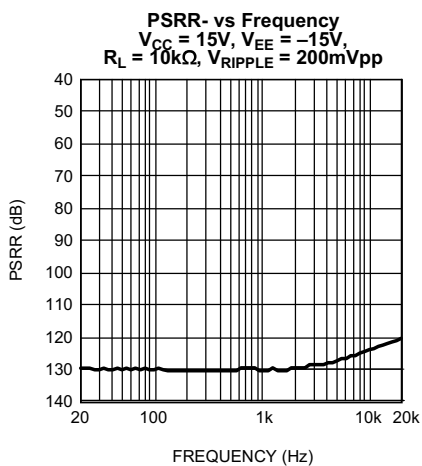


Figure 57.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

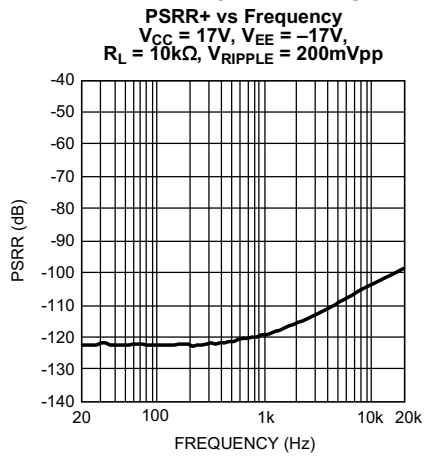


Figure 58.

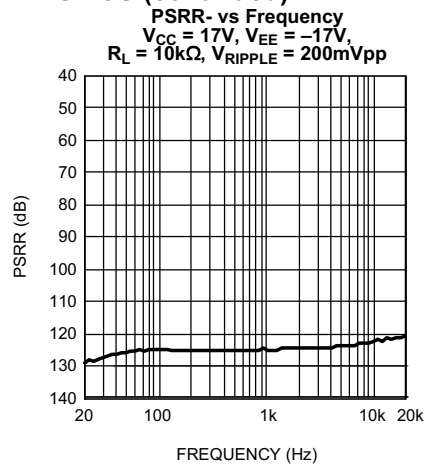


Figure 59.

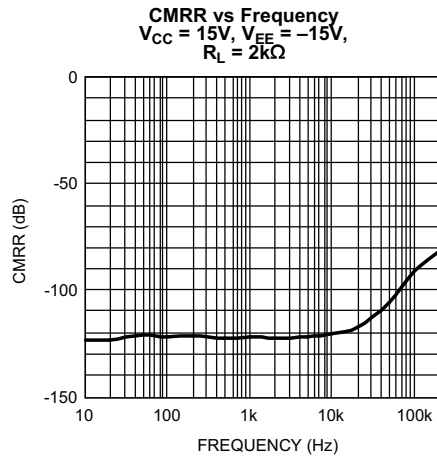


Figure 60.

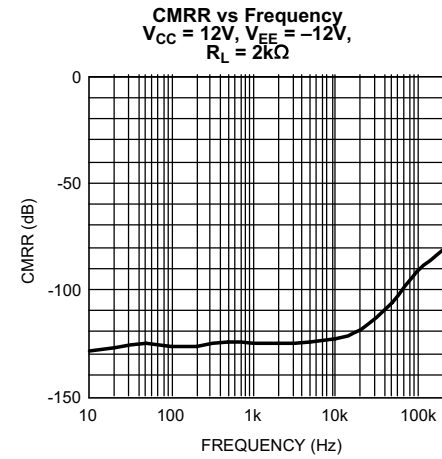


Figure 61.

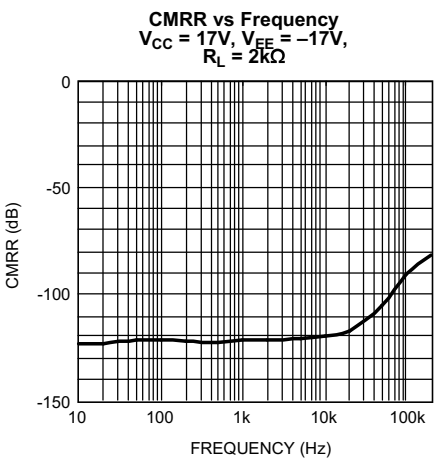


Figure 62.

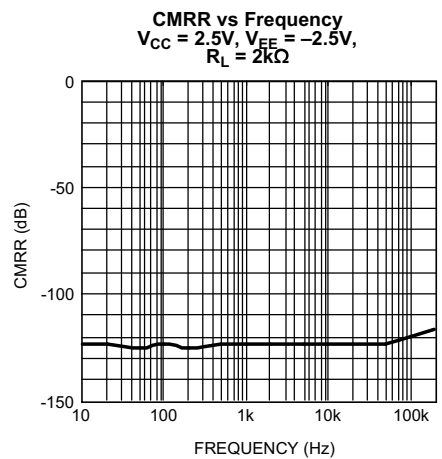


Figure 63.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

CMRR vs Frequency
 $V_{CC} = 15V, V_{EE} = -15V,$
 $R_L = 600\Omega$

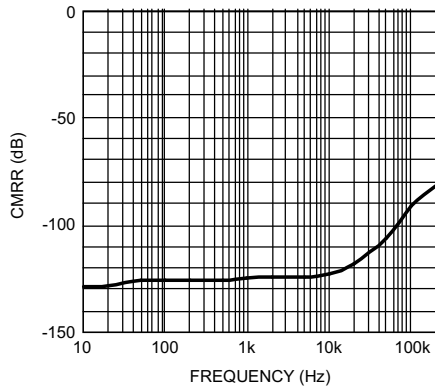


Figure 64.

CMRR vs Frequency
 $V_{CC} = 12V, V_{EE} = -12V,$
 $R_L = 600\Omega$

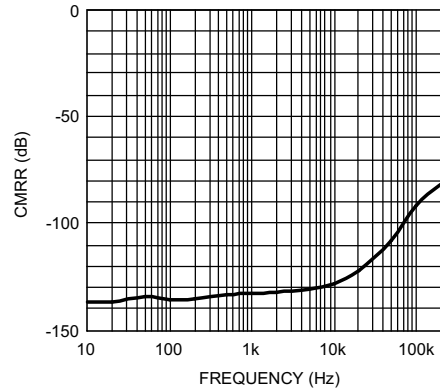


Figure 65.

CMRR vs Frequency
 $V_{CC} = 17V, V_{EE} = -17V,$
 $R_L = 600\Omega$

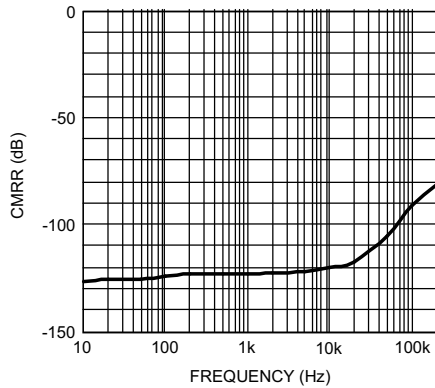


Figure 66.

CMRR vs Frequency
 $V_{CC} = 2.5V, V_{EE} = -2.5V,$
 $R_L = 600\Omega$

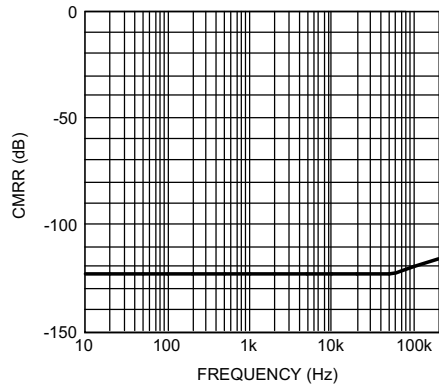


Figure 67.

CMRR vs Frequency
 $V_{CC} = 15V, V_{EE} = -15V,$
 $R_L = 10k\Omega$

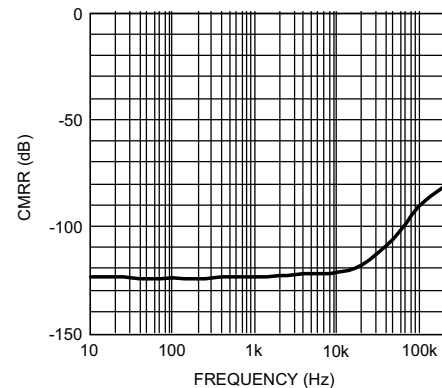


Figure 68.

CMRR vs Frequency
 $V_{CC} = 12V, V_{EE} = -12V,$
 $R_L = 10k\Omega$

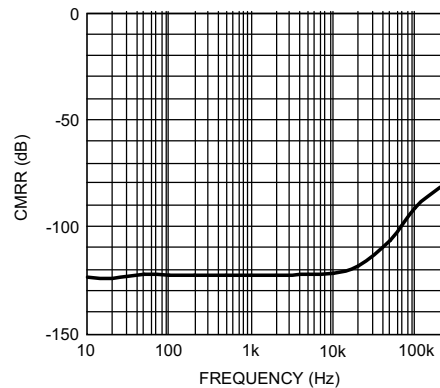


Figure 69.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

CMRR vs Frequency
 $V_{CC} = 17V, V_{EE} = -17V,$
 $R_L = 10k\Omega$

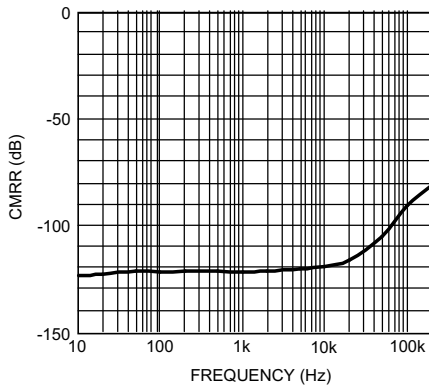


Figure 70.

CMRR vs Frequency
 $V_{CC} = 2.5V, V_{EE} = -2.5V,$
 $R_L = 10k\Omega$

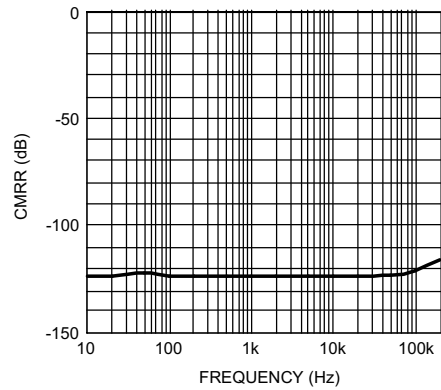


Figure 71.

Output Voltage vs Supply Voltage
 $R_L = 2k\Omega, THD+N = 1\%$

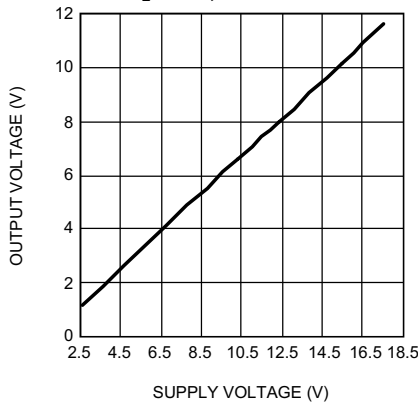


Figure 72.

Output Voltage vs Supply Voltage
 $R_L = 600\Omega, THD+N = 1\%$

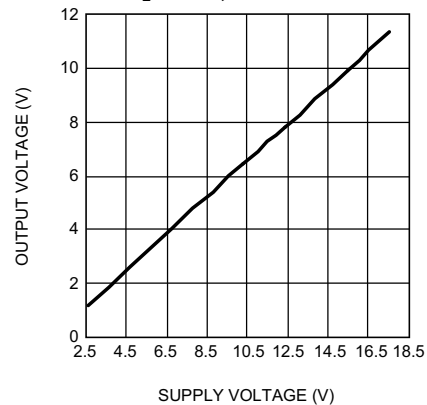


Figure 73.

Output Voltage vs Supply Voltage
 $R_L = 10k\Omega, THD+N = 1\%$

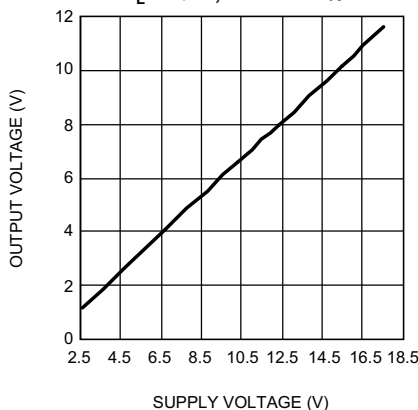


Figure 74.

Output Voltage vs Load Resistance
 $V_{CC} = 15V, V_{EE} = -15V, THD+N = 1\%$

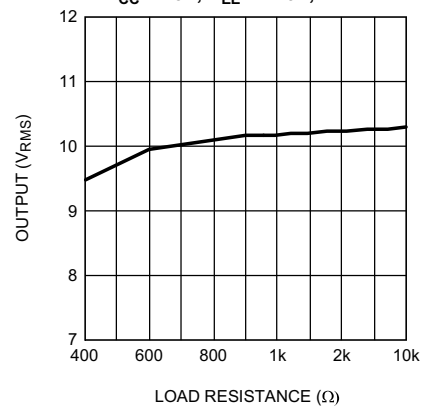


Figure 75.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

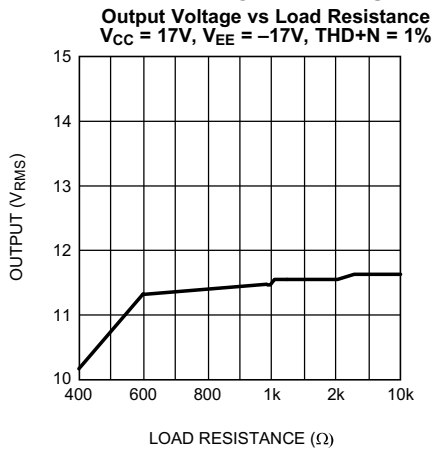


Figure 76.

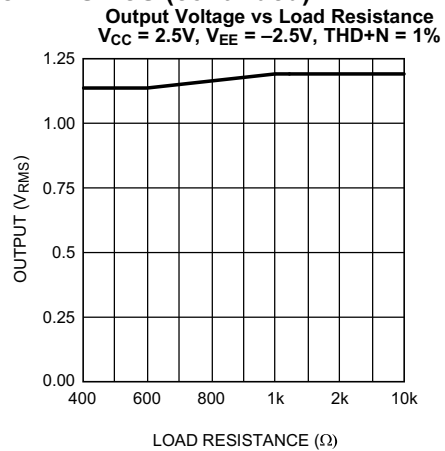


Figure 77.

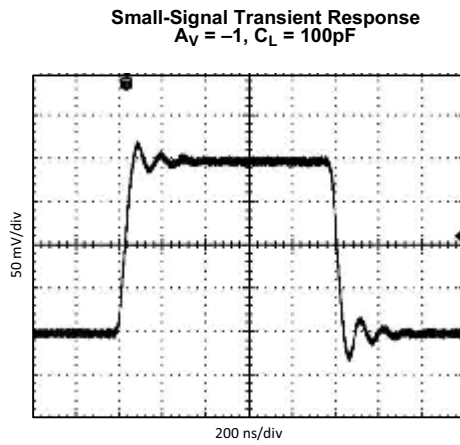


Figure 78.

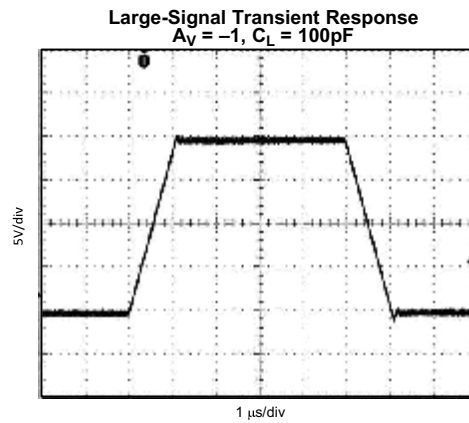
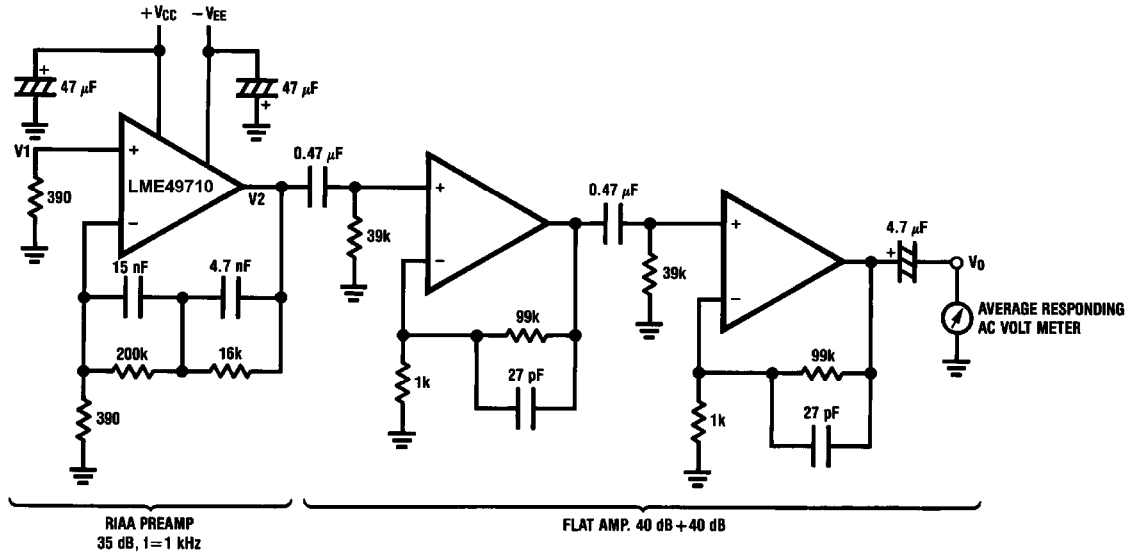


Figure 79.

NOISE MEASUREMENT CIRCUIT



A. Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.

Figure 80. Total Gain: 115 dB at $f = 1$ kHz
 Input Referred Noise Voltage: $e_n = V_o / 560,000$ (V)

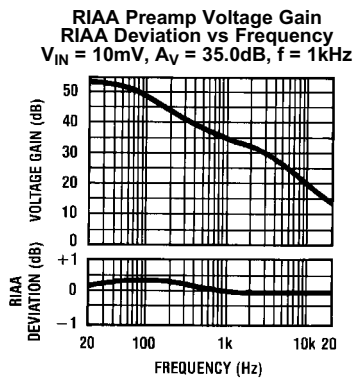


Figure 81.

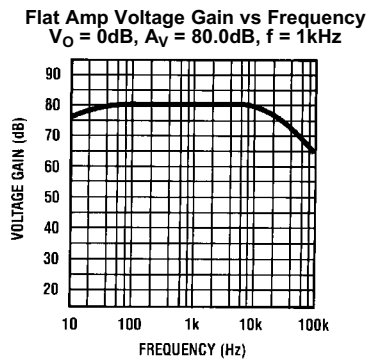


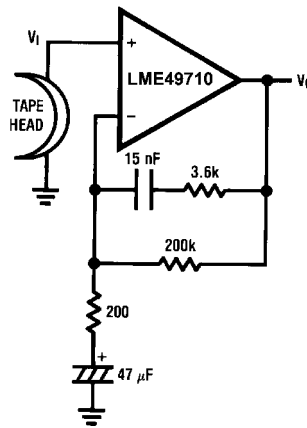
Figure 82.

APPLICATION HINTS

The LME49710 is a high-speed op amp with excellent phase margin and stability. Capacitive loads up to 100pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 100pF must be isolated from the output. The most straight forward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.

TYPICAL APPLICATIONS



$A_v = 34.5$
 $F = 1 \text{ kHz}$
 $E_n = 0.38 \mu\text{V}$
 A Weighted

Figure 83. NAB Preamp

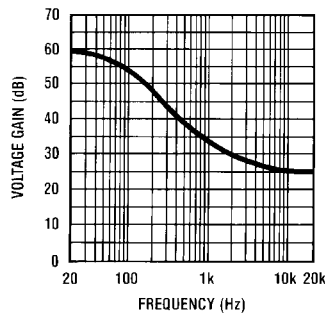
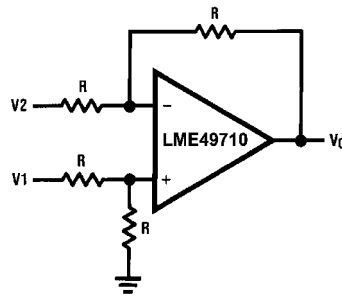
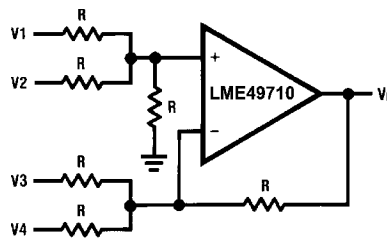


Figure 84. NAB Preamp Voltage Gain vs Frequency
 $V_{IN} = 10\text{mV}$, 34.5dB , $f = 1\text{kHz}$



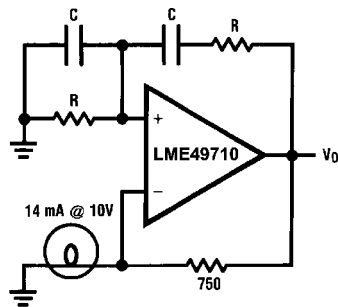
$$V_O = V_1 - V_2$$

Figure 85. Balanced to Single Ended Converter



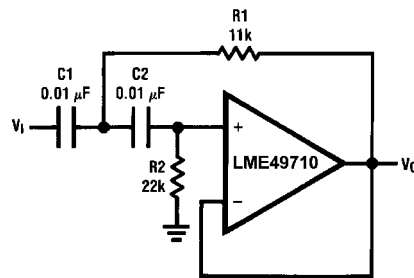
$$V_O = V_1 + V_2 - V_3 - V_4$$

Figure 86. Adder/Subtractor



$$f_o = \frac{1}{2\pi RC}$$

Figure 87. Sine Wave Oscillator



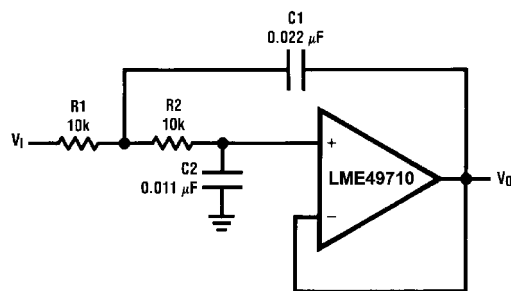
if $C1 = C2 = C$

$$R1 = \frac{\sqrt{2}}{2\omega_0 C}$$

$$R2 = 2 \cdot R1$$

Illustration is $f_0 = 1 \text{ kHz}$

Figure 88. Second-Order High-Pass Filter (Butterworth)



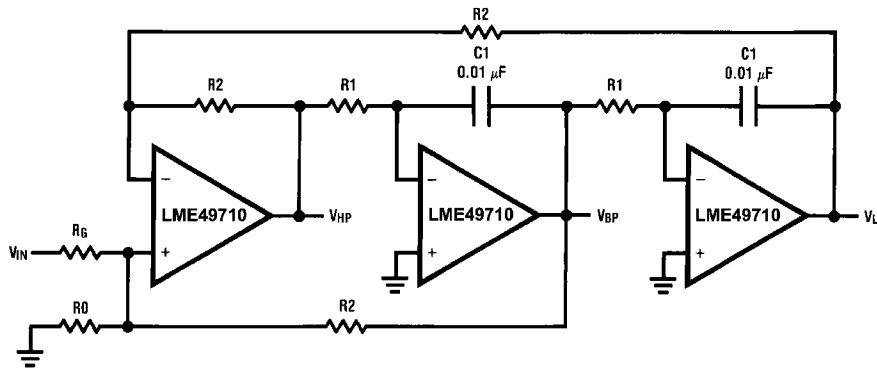
if $R1 = R2 = R$

$$C1 = \frac{\sqrt{2}}{\omega_0 R}$$

$$C2 = \frac{C1}{2}$$

Illustration is $f_0 = 1 \text{ kHz}$

Figure 89. Second-Order Low-Pass Filter (Butterworth)



$$f_0 = \frac{1}{2\pi C1 R1}, Q = \frac{1}{2} \left(1 + \frac{R2}{R0} + \frac{R2}{RG} \right), A_{BP} = Q A_{LP} = Q A_{LH} = \frac{R2}{RG}$$

Figure 90. State Variable Filter

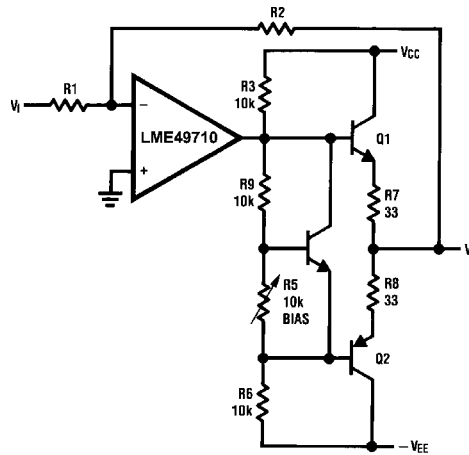
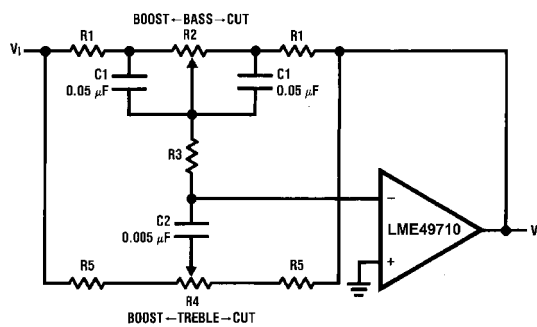


Figure 91. Line Driver



$$f_L = \frac{1}{2\pi R2 C1}, f_{LB} = \frac{1}{2\pi R1 C1}$$

$$f_H = \frac{1}{2\pi R5 C2}, f_{HB} = \frac{1}{2\pi (R1 + R5 + 2R3) C2}$$

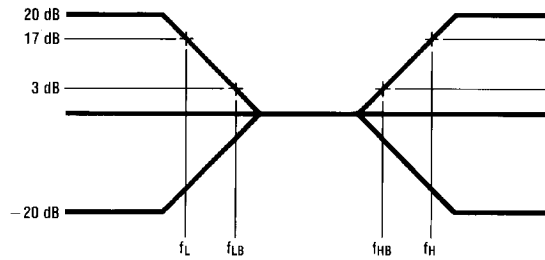
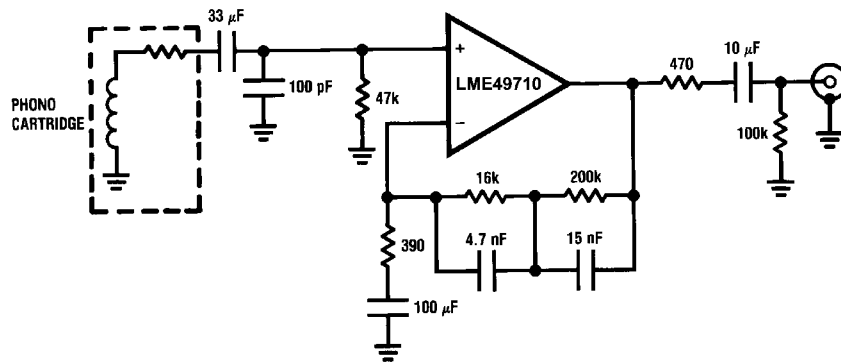
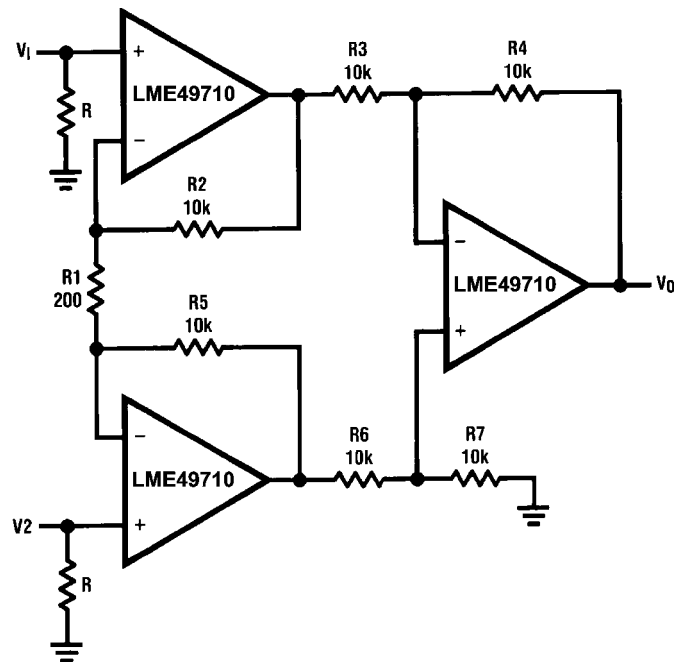


Figure 92. Tone Control



$A_v = 35 \text{ dB}$
 $E_n = 0.33 \mu\text{V}$
 $S/N = 90 \text{ dB}$
 $f = 1 \text{ kHz}$
 A Weighted
 A Weighted, $V_{IN} = 10 \text{ mV}$
 @ $f = 1 \text{ kHz}$

Figure 93. RIAA Preamp



If $R2 = R5$, $R3 = R6$, $R4 = R7$

$$V_0 = \left(1 + \frac{2R2}{R1}\right) \frac{R4}{R3} (V2 - V1)$$

Illustration is:

$$V_0 = 101(V2 - V1)$$

Figure 94. Balanced Input Mic Amp

APPLICATION INFORMATION

DISTORTION MEASUREMENTS

The vanishingly low-residual distortion produced by LME49710 is below the capabilities of all commercially available equipment. This makes distortion measurements just slightly more difficult than simply connecting a distortion meter to the amplifier's inputs and outputs. The solution, however, is quite simple: an additional resistor. Adding this resistor extends the resolution of the distortion measurement equipment.

The LME49710's low-residual distortion is an input referred internal error. As shown in [Figure 95](#), adding the 10Ω resistor connected between the amplifier's inverting and non-inverting inputs changes the amplifier's noise gain. The result is that the error signal (distortion) is amplified by a factor of 101. Although the amplifier's closed-loop gain is unaltered, the feedback available to correct distortion errors is reduced by 101, which means that measurement resolution increases by 101. To ensure minimum effects on distortion measurements, keep the value of R1 low as shown in [Figure 95](#).

This technique is verified by duplicating the measurements with high closed-loop gain and/or making the measurements at high frequencies. Doing so produces distortion components that are within the measurement equipment's capabilities. This datasheet's THD+N and IMD values were generated using the above described circuit connected to an Audio Precision System Two Cascade.

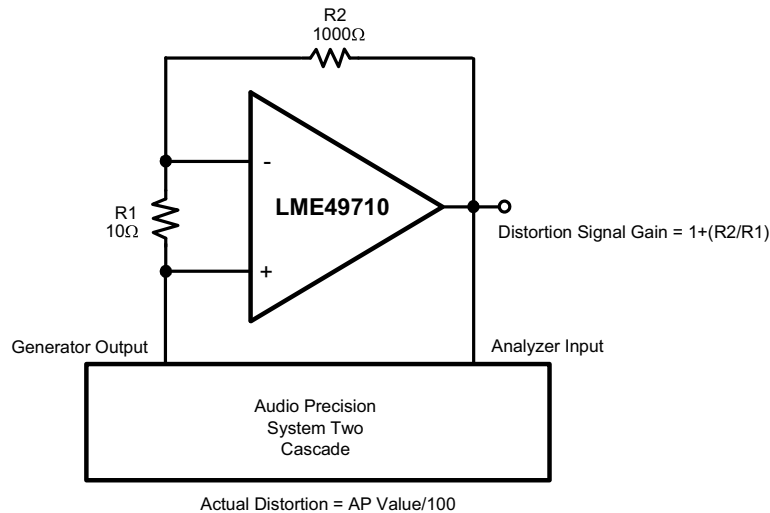


Figure 95. THD+N and IMD Distortion Test Circuit