

## LME49720 Dual High Performance, High Fidelity Audio Operational Amplifier

 Check for Samples: [LME49720](#)

### FEATURES

- Easily Drives 600Ω Loads
- Optimized for Superior Audio Signal Fidelity
- Output Short Circuit Protection
- PSRR and CMRR Exceed 120dB (typ)
- SOIC, PDIP, TO-99 Metal Can Packages

### APPLICATIONS

- Ultra High Quality Audio Amplification
- High Fidelity Preamplifiers
- High Fidelity Multimedia
- State of the Art Phono Pre Amps
- High Performance Professional Audio
- High Fidelity Equalization and Crossover Networks
- High Performance Line Drivers
- High Performance Line Receivers
- High Fidelity Active Filters

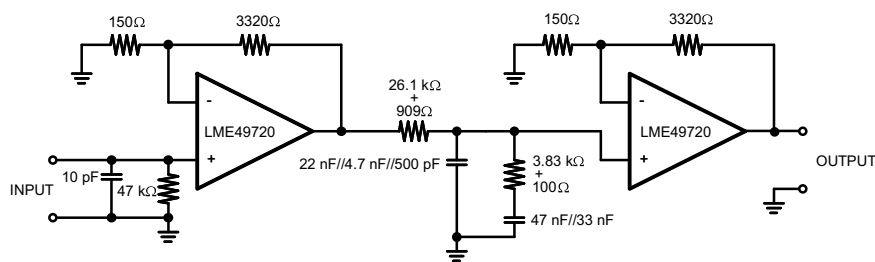
### KEY SPECIFICATIONS

- Power Supply Voltage Range: ±2.5 to ±17V
- THD+N ( $A_V = 1$ ,  $V_{OUT} = 3V_{RMS}$ ,  $f_{IN} = 1kHz$ ):
  - $R_L = 2k\Omega$ : 0.00003% (typ)
  - $R_L = 600\Omega$ : 0.00003% (typ)
- Input Noise Density: 2.7nV/√Hz (typ)
- Slew Rate: ±20V/μs (typ)
- Gain Bandwidth Product: 55MHz (typ)
- Open Loop Gain ( $R_L = 600\Omega$ ): 140dB (typ)
- Input Bias Current: 10nA (typ)
- Input Offset Voltage: 0.1mV (typ)
- DC Gain Linearity Error: 0.000009%

### DESCRIPTION

The LME49720 is part of the ultra-low distortion, low noise, high slew rate operational amplifier series optimized and fully specified for high performance, high fidelity applications. Combining advanced leading-edge process technology with state-of-the-art circuit design, the LME49720 audio operational amplifiers deliver superior audio signal amplification for outstanding audio performance. The LME49720 combines extremely low voltage noise density (2.7nV/√Hz) with vanishingly low THD+N (0.00003%) to easily satisfy the most demanding audio applications.

### TYPICAL APPLICATION



Note: 1% metal film resistors, 5% polypropylene capacitors

**Figure 1. Passively Equalized RIAA Phono Preamplifier**


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**DESCRIPTION (CONTINUED)**

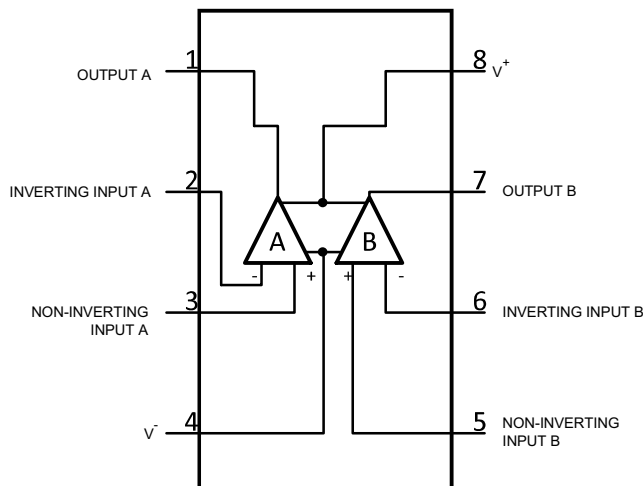
To ensure that the most challenging loads are driven without compromise, the LME49720 has a high slew rate of  $\pm 20V/\mu s$  and an output current capability of  $\pm 26mA$ . Further, dynamic range is maximized by an output stage that drives  $2k\Omega$  loads to within 1V of either power supply voltage and to within 1.4V when driving  $600\Omega$  loads.

The LME49720's outstanding CMRR (120dB), PSRR (120dB), and  $V_{OS}$  (0.1mV) give the amplifier excellent operational amplifier DC performance.

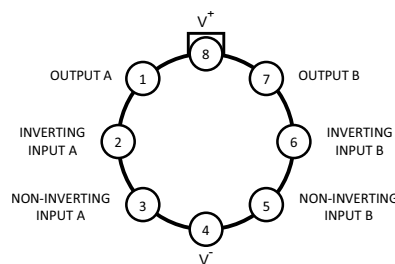
The LME49720 has a wide supply range of  $\pm 2.5V$  to  $\pm 17V$ . Over this supply range the LME49720's input circuitry maintains excellent common-mode and power supply rejection, as well as maintaining its low input bias current. The LME49720 is unity gain stable. This Audio Operational Amplifier achieves outstanding AC performance while driving complex loads with values as high as  $100pF$ .

The LME49720 is available in 8-lead narrow body SOIC, 8-lead PDIP, and 8-lead TO-99. Demonstration boards are available for each package.

**Connection Diagrams**



**Figure 2. 8-Pin SOIC or PDIP  
See D or P Package**



**Figure 3. 8-Lead TO-99  
See LMC Package**



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

**ABSOLUTE MAXIMUM RATINGS** <sup>(1)(2)(3)</sup>

Power Supply Voltage	$(V_S = V^+ - V^-)$	36V
Storage Temperature		-65°C to 150°C
Input Voltage		$(V^-) - 0.7V$ to $(V^+) + 0.7V$
Output Short Circuit <sup>(4)</sup>		Continuous
Power Dissipation		Internally Limited
ESD Susceptibility <sup>(5)</sup>		2000V
ESD Susceptibility <sup>(6)</sup>	Pins 1, 4, 7 and 8	200V
	Pins 2, 3, 5 and 6	100V
Junction Temperature		150°C
Thermal Resistance	$\theta_{JA}$ (SOIC)	145°C/W
	$\theta_{JA}$ (PDIP)	102°C/W
	$\theta_{JA}$ (TO-99)	150°C/W
	$\theta_{JC}$ (TO-99)	35°C/W
Temperature Range	$T_{MIN} \leq T_A \leq T_{MAX}$	-40°C $\leq T_A \leq$ 85°C
Supply Voltage Range		$\pm 2.5V \leq V_S \leq \pm 17V$

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.
- (2) Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. For ensured specifications and test conditions, see Electrical Characteristics. The ensured specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.
- (3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (4) Amplifier output connected to GND, any number of amplifiers within a package.
- (5) Human body model, 100pF discharged through a 1.5k $\Omega$  resistor.
- (6) Machine Model ESD test is covered by specification EIAJ IC-121-1981. A 200pF cap is charged to the specified voltage and then discharged directly into the IC with no external series resistor (resistance of discharge path must be under 50 $\Omega$ ).

**ELECTRICAL CHARACTERISTICS FOR THE LME49720** <sup>(1)(2)</sup>

The following specifications apply for  $V_S = \pm 15V$ ,  $R_L = 2k\Omega$ ,  $f_{IN} = 1kHz$ , and  $T_A = 25^\circ C$ , unless otherwise specified.

Symbol	Parameter	Conditions	LME49720		Units (Limits)
			Typical <sup>(3)</sup>	Limit <sup>(4)</sup>	
THD+N	Total Harmonic Distortion + Noise	$A_V = 1$ , $V_{OUT} = 3V_{RMS}$ $R_L = 2k\Omega$ $R_L = 600\Omega$	0.00003 0.00003	0.00009	% (max)
IMD	Intermodulation Distortion	$A_V = 1$ , $V_{OUT} = 3V_{RMS}$ Two-tone, 60Hz & 7kHz 4:1	0.00005		%
GBWP	Gain Bandwidth Product		55	45	MHz (min)
SR	Slew Rate		$\pm 20$	$\pm 15$	V/ $\mu s$ (min)
FPBW	Full Power Bandwidth	$V_{OUT} = 1V_{P-P}$ , -3dB referenced to output magnitude at $f = 1kHz$	10		MHz
$t_s$	Settling time	$A_V = -1$ , 10V step, $C_L = 100pF$ 0.1% error range	1.2		$\mu s$
$e_n$	Equivalent Input Noise Voltage	$f_{BW} = 20Hz$ to 20kHz	0.34	0.65	$\mu V_{RMS}$ (max)
	Equivalent Input Noise Density	$f = 1kHz$ $f = 10Hz$	2.7 6.4	4.7	nV/ $\sqrt{Hz}$ (max)
$i_n$	Current Noise Density	$f = 1kHz$	1.6		pA/ $\sqrt{Hz}$
		$f = 10Hz$	3.1		
$V_{OS}$	Offset Voltage		$\pm 0.1$	$\pm 0.7$	mV (max)

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.
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- (3) Typical specifications are specified at +25°C and represent the most likely parametric norm.
- (4) Tested limits are ensured to AOQL (Average Outgoing Quality Level).

**ELECTRICAL CHARACTERISTICS FOR THE LME49720 <sup>(1)(2)</sup> (continued)**

The following specifications apply for  $V_S = \pm 15V$ ,  $R_L = 2k\Omega$ ,  $f_{IN} = 1kHz$ , and  $T_A = 25^\circ C$ , unless otherwise specified.

Symbol	Parameter	Conditions	LME49720		Units (Limits)
			Typical <sup>(3)</sup>	Limit <sup>(4)</sup>	
$\Delta V_{OS}/\Delta Temp$	Average Input Offset Voltage Drift vs Temperature	$-40^\circ C \leq T_A \leq 85^\circ C$	0.2		$\mu V/^\circ C$
PSRR	Average Input Offset Voltage Shift vs Power Supply Voltage	$\Delta V_S = 20V$ <sup>(5)</sup>	120	110	dB (min)
ISO <sub>CH-CH</sub>	Channel-to-Channel Isolation	$f_{IN} = 1kHz$ $f_{IN} = 20kHz$	118 112		dB
$I_B$	Input Bias Current	$V_{CM} = 0V$	10	72	nA (max)
$\Delta I_{OS}/\Delta Temp$	Input Bias Current Drift vs Temperature	$-40^\circ C \leq T_A \leq 85^\circ C$	0.1		nA/°C
$I_{OS}$	Input Offset Current	$V_{CM} = 0V$	11	65	nA (max)
$V_{IN-CM}$	Common-Mode Input Voltage Range		+14.1 -13.9	(V+) - 2.0 (V-) + 2.0	V (min)
CMRR	Common-Mode Rejection	$-10V < V_{cm} < 10V$	120	110	dB (min)
$Z_{IN}$	Differential Input Impedance		30		k $\Omega$
	Common Mode Input Impedance	$-10V < V_{cm} < 10V$	1000		M $\Omega$
$A_{VOL}$	Open Loop Voltage Gain	$-10V < V_{out} < 10V$ , $R_L = 600\Omega$	140	125	dB (min)
		$-10V < V_{out} < 10V$ , $R_L = 2k\Omega$	140		
		$-10V < V_{out} < 10V$ , $R_L = 10k\Omega$	140		
$V_{OUTMAX}$	Maximum Output Voltage Swing	$R_L = 600\Omega$	$\pm 13.6$	$\pm 12.5$	V (min)
		$R_L = 2k\Omega$	$\pm 14.0$		
		$R_L = 10k\Omega$	$\pm 14.1$		
$I_{OUT}$	Output Current	$R_L = 600\Omega$ , $V_S = \pm 17V$	$\pm 26$	$\pm 23$	mA (min)
$I_{OUT-CC}$	Instantaneous Short Circuit Current		+53 -42		mA
$R_{OUT}$	Output Impedance	$f_{IN} = 10kHz$ Closed-Loop Open-Loop	0.01 13		$\Omega$
$C_{LOAD}$	Capacitive Load Drive Overshoot	100pF	16		%
$I_S$	Total Quiescent Current	$I_{OUT} = 0mA$	10	12	mA (max)

(5) PSRR is measured as follows:  $V_{OS}$  is measured at two supply voltages,  $\pm 5V$  and  $\pm 15V$ .  $PSRR = |20\log(\Delta V_{OS}/\Delta V_S)|$ .

TYPICAL PERFORMANCE CHARACTERISTICS

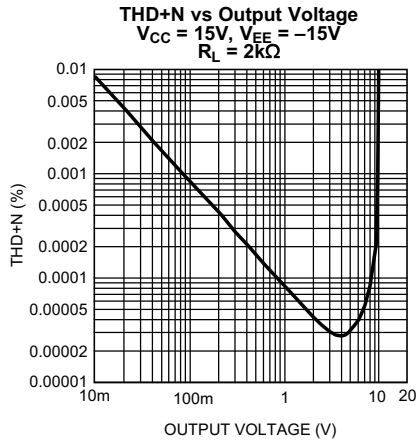


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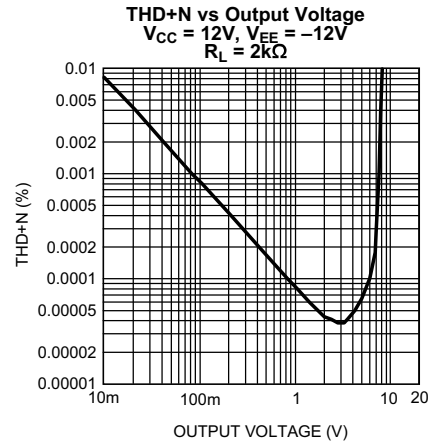


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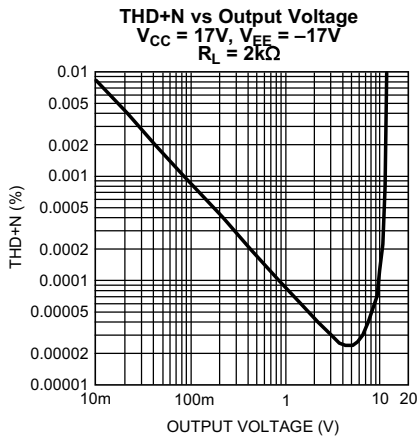


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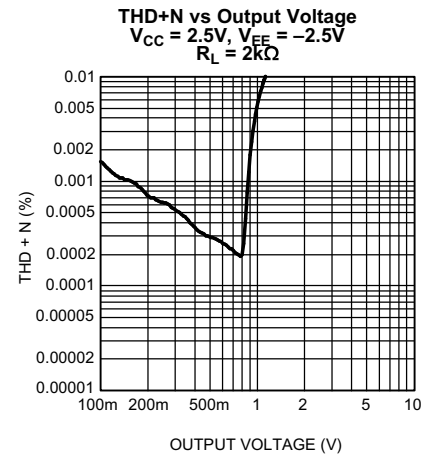


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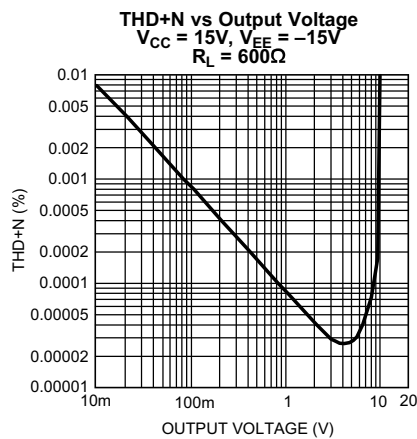


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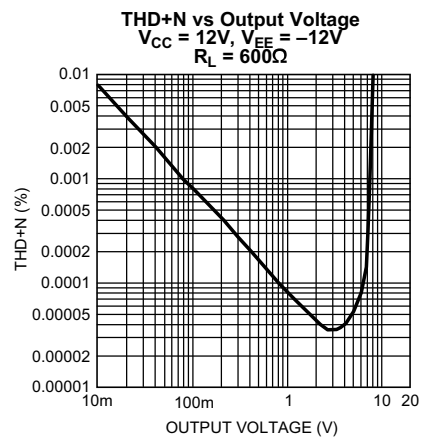


Figure 9.

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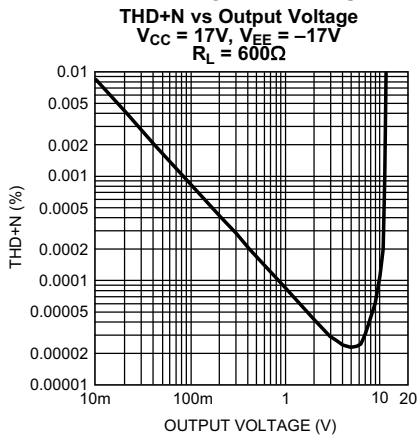


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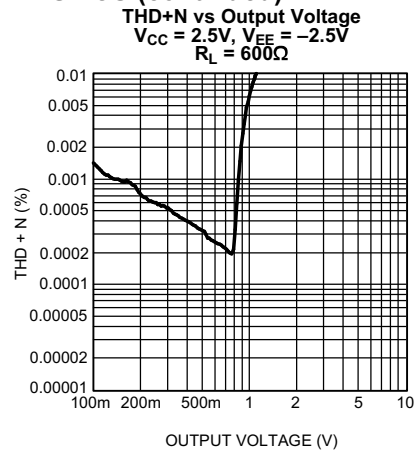


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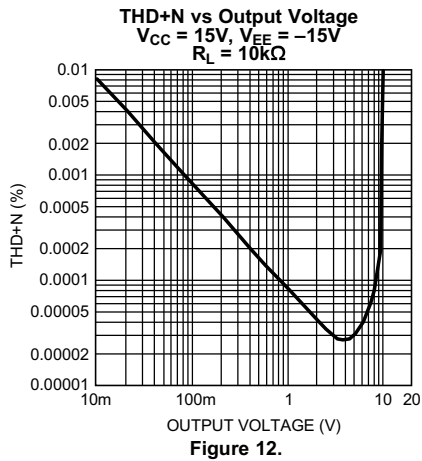


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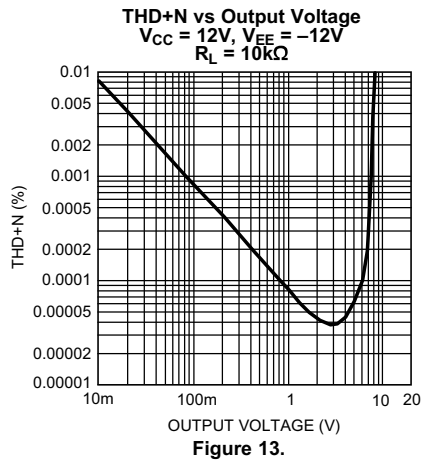


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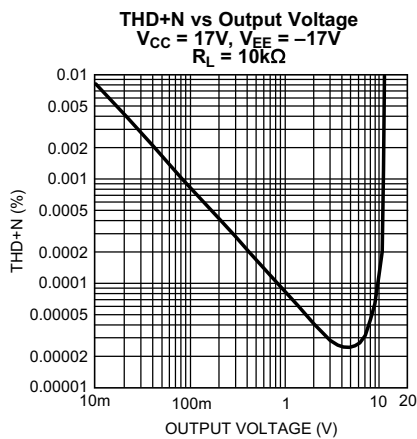


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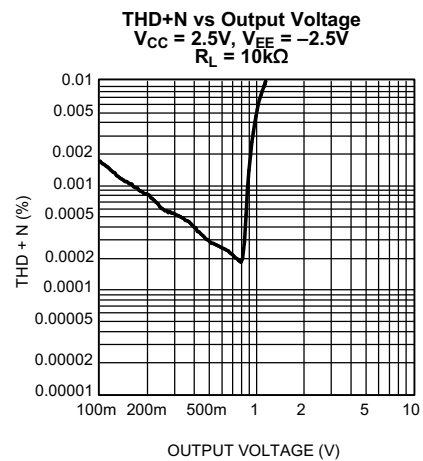


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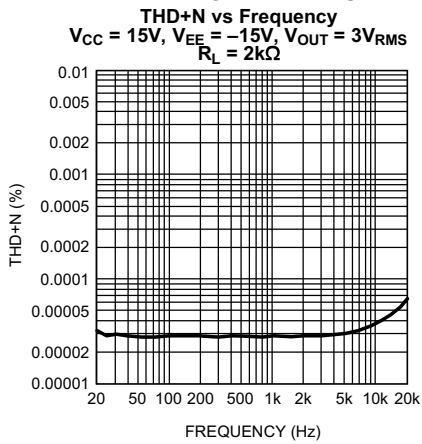


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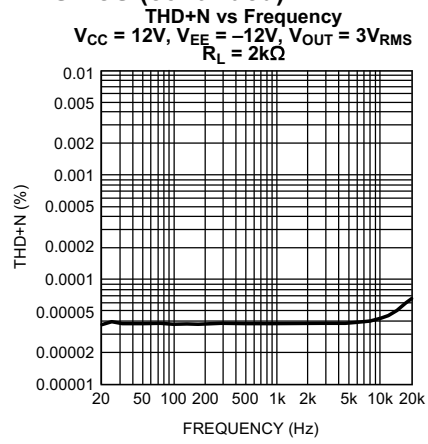


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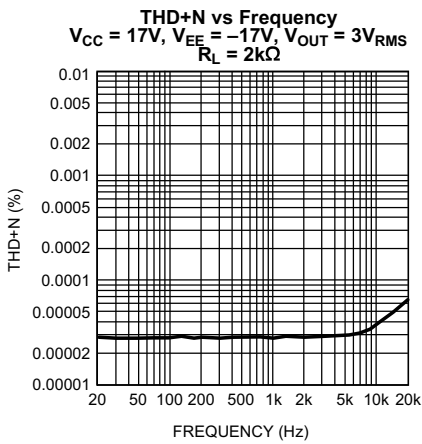


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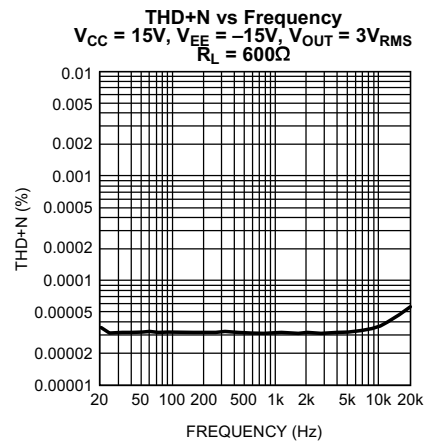


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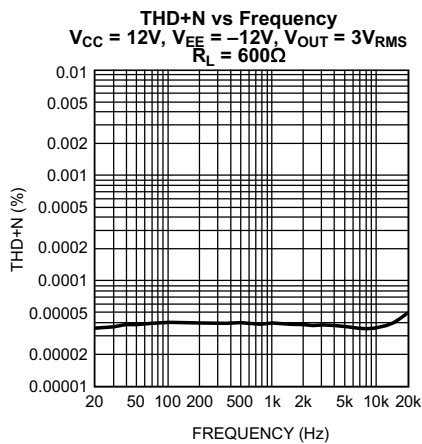


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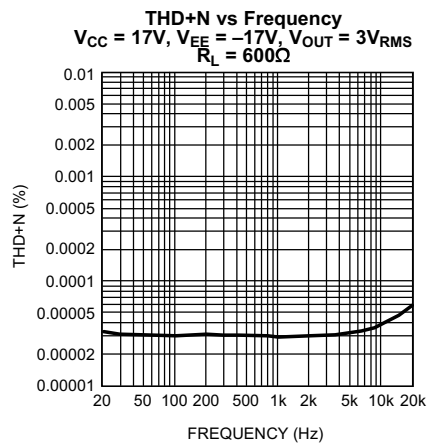


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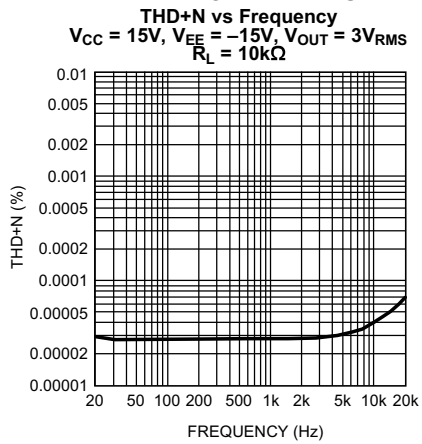


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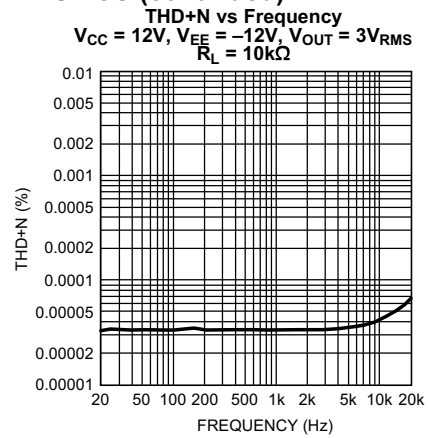


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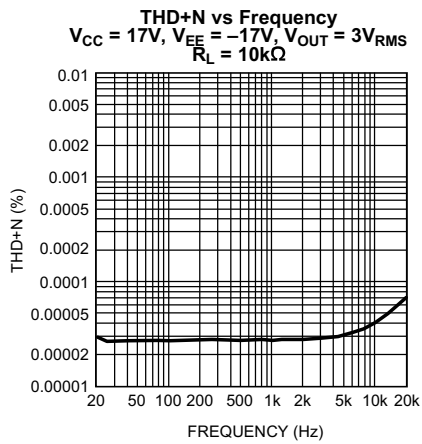


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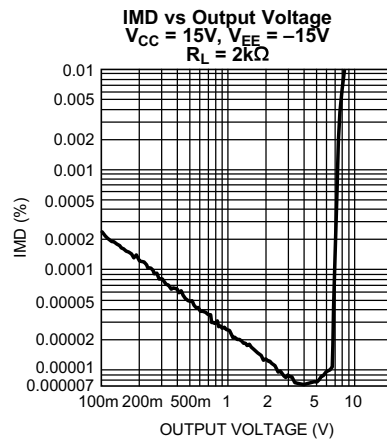


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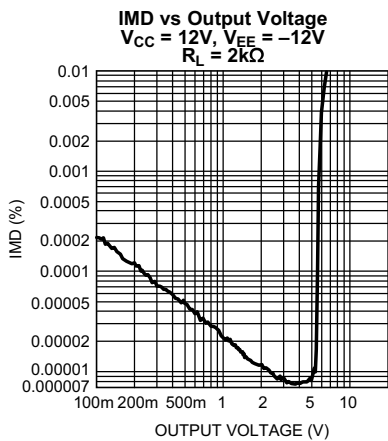


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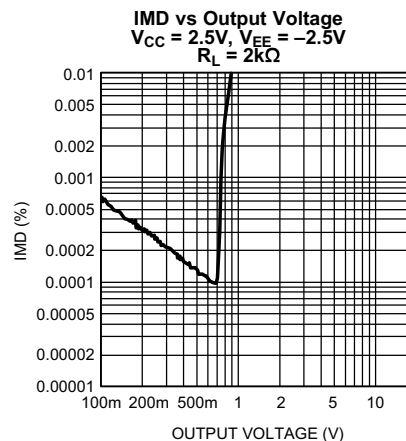


Figure 27.



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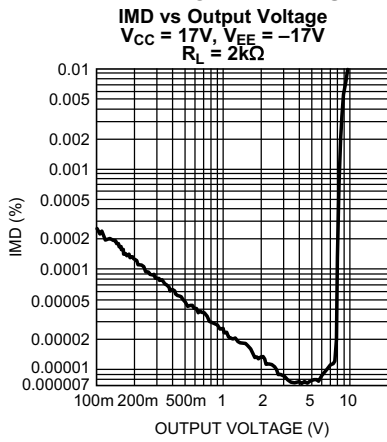


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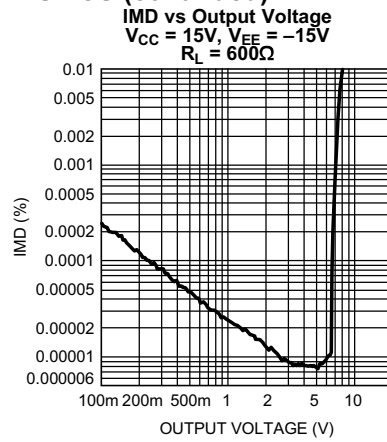


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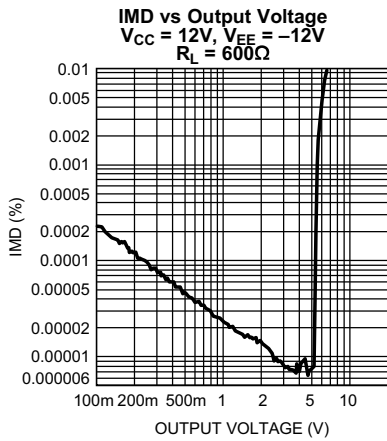


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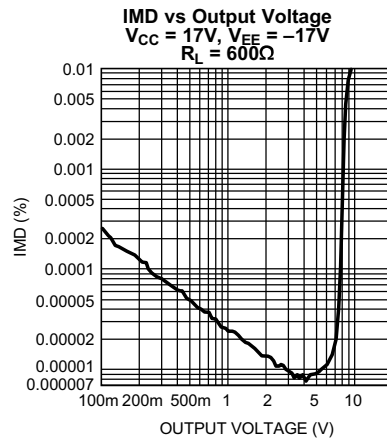


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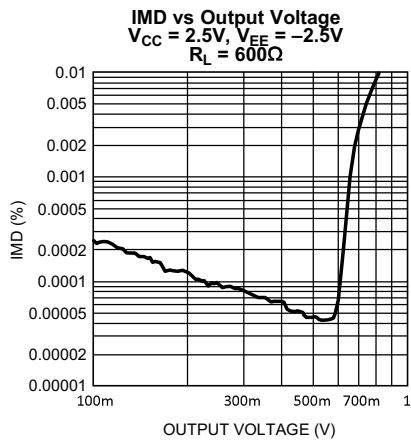


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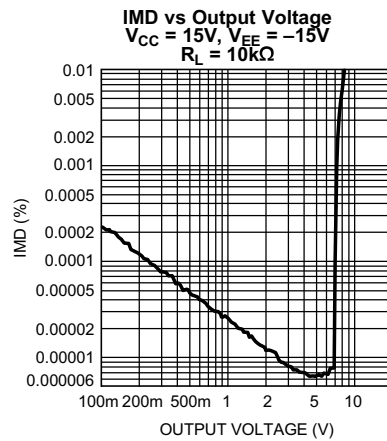


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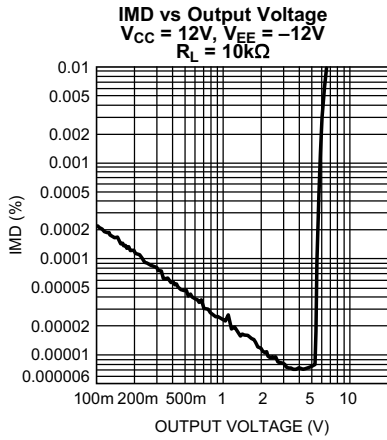


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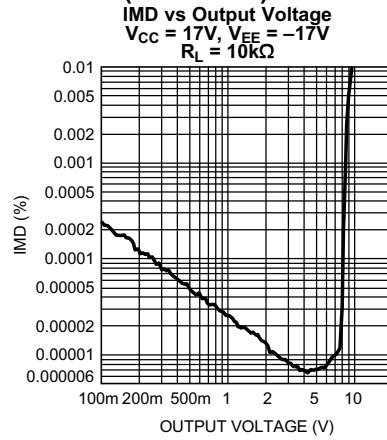


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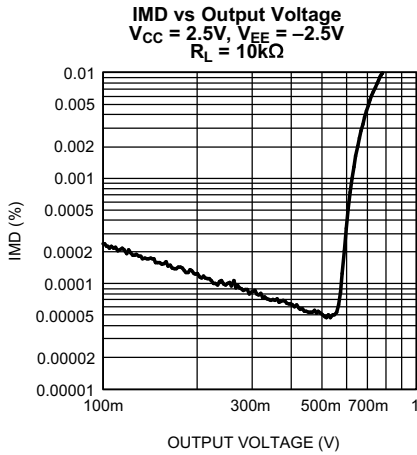


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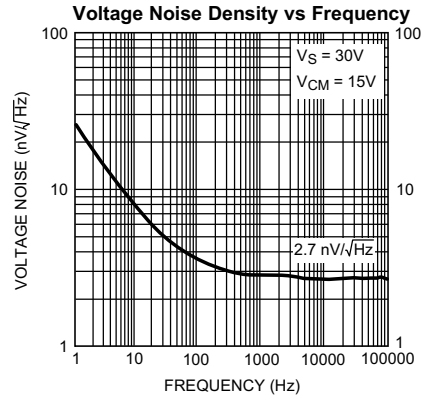


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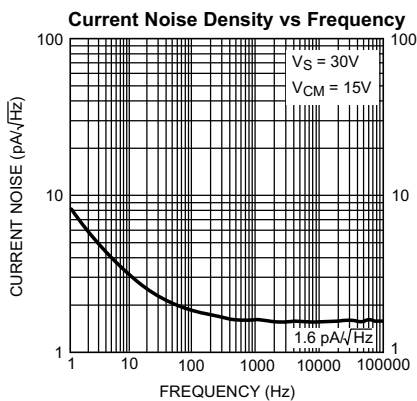


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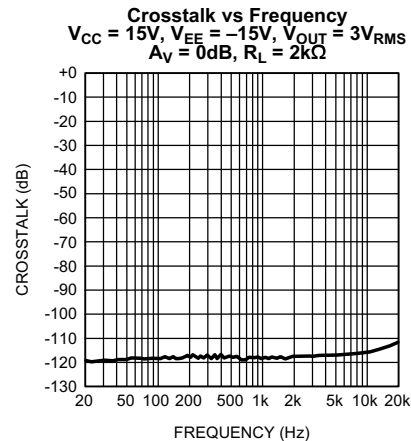


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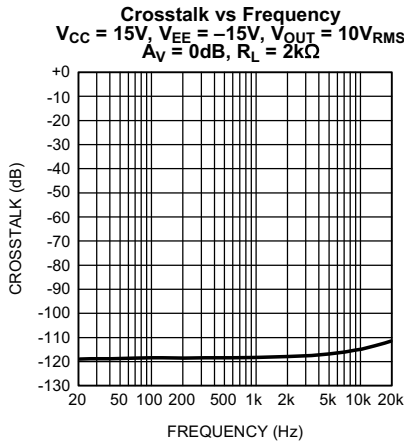


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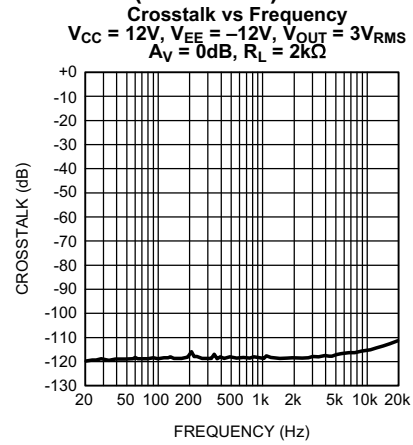


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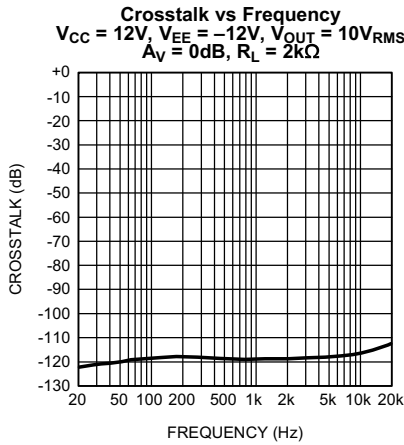


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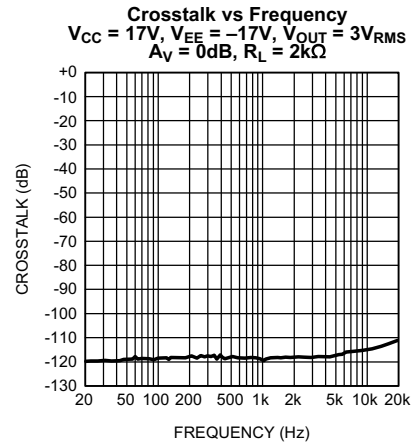


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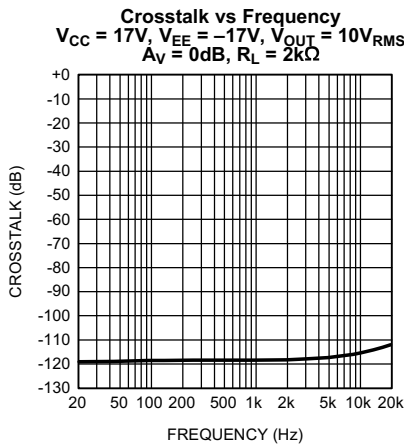


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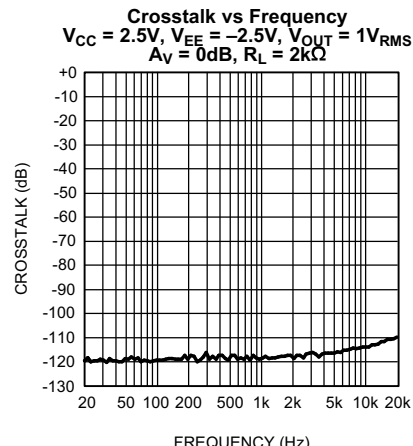


Figure 45.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

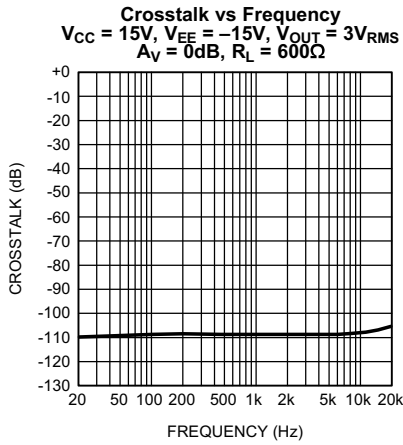


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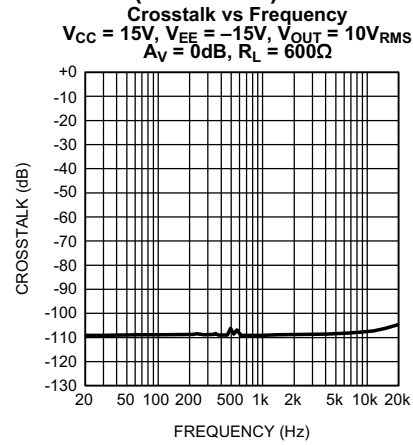


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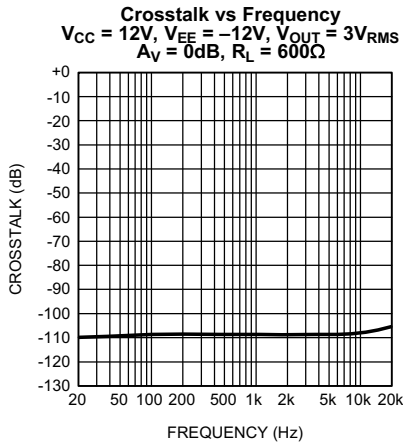


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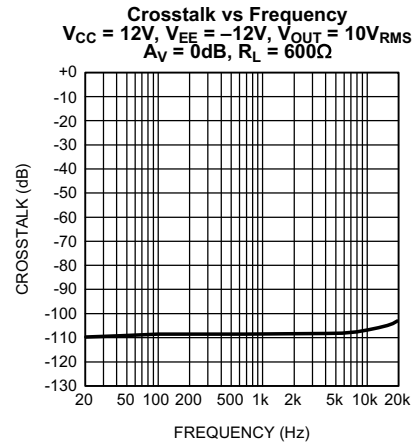


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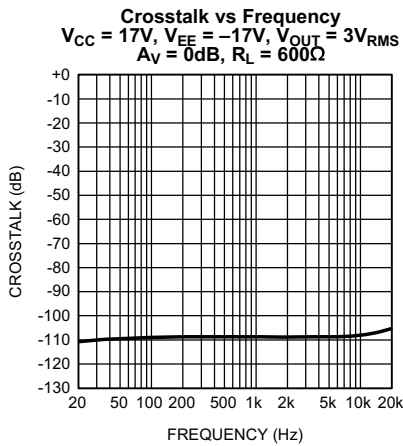


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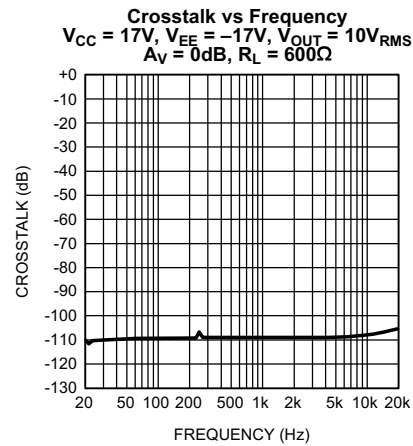


Figure 51.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

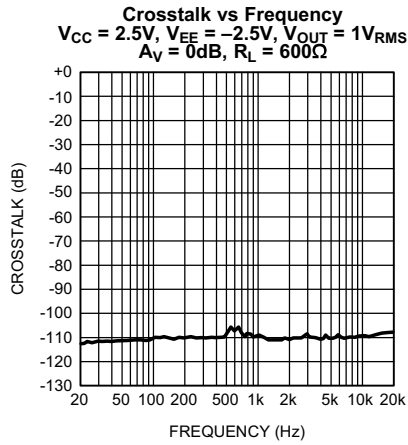


Figure 52.

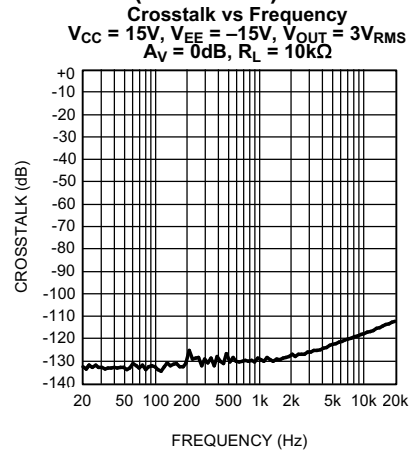


Figure 53.

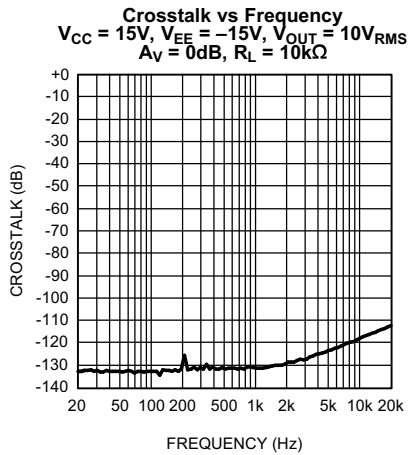


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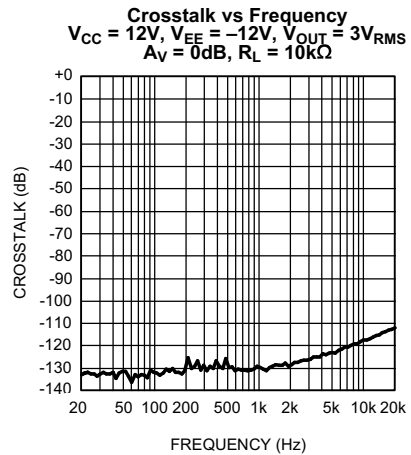


Figure 55.

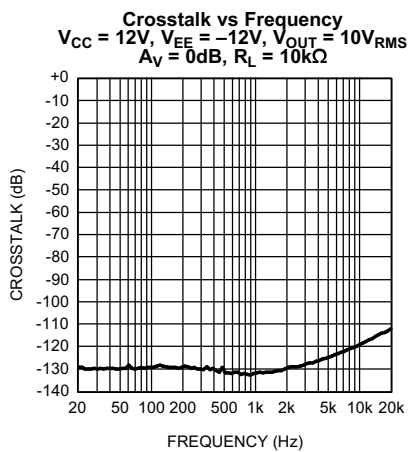


Figure 56.

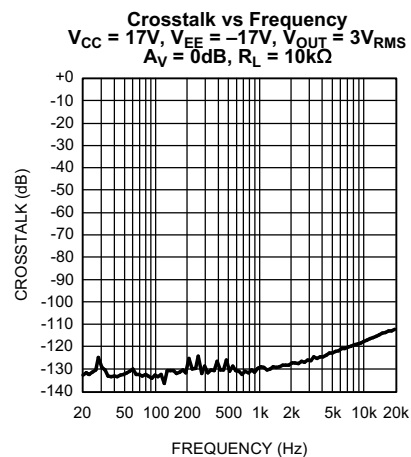


Figure 57.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

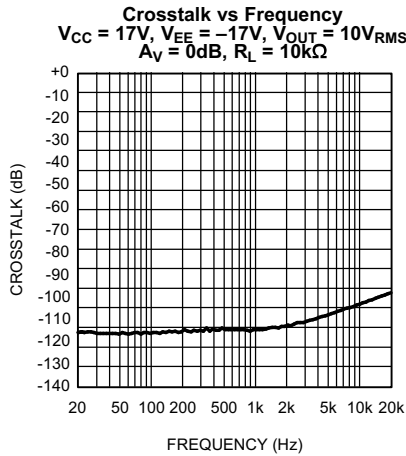


Figure 58.

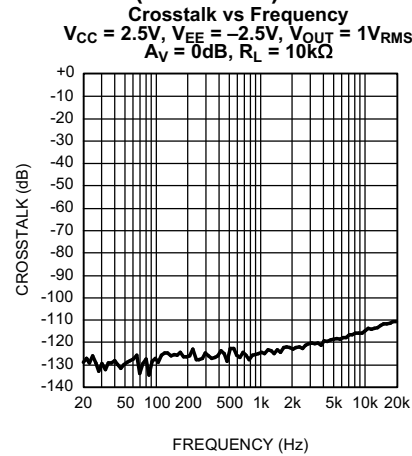


Figure 59.

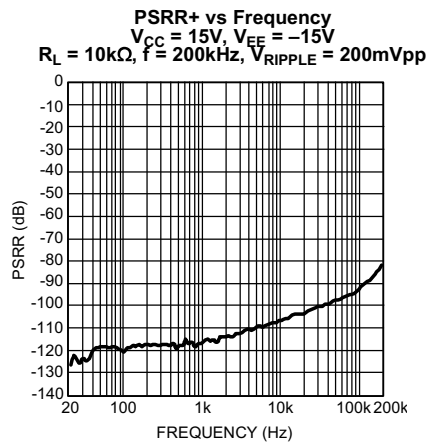


Figure 60.

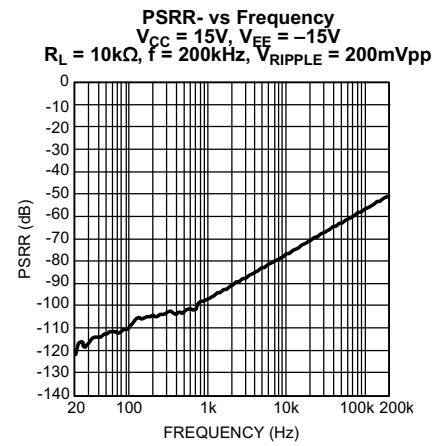


Figure 61.

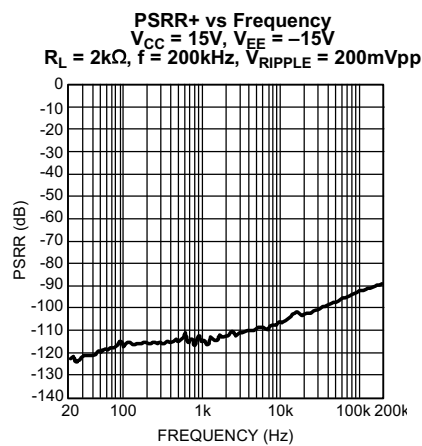


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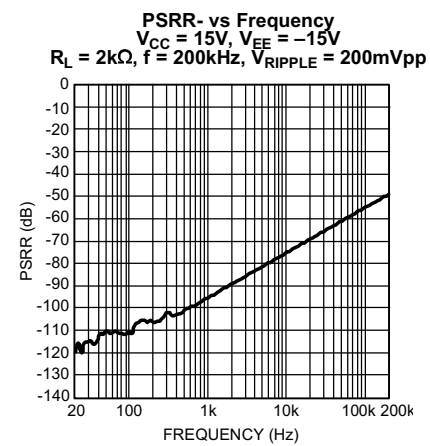


Figure 63.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

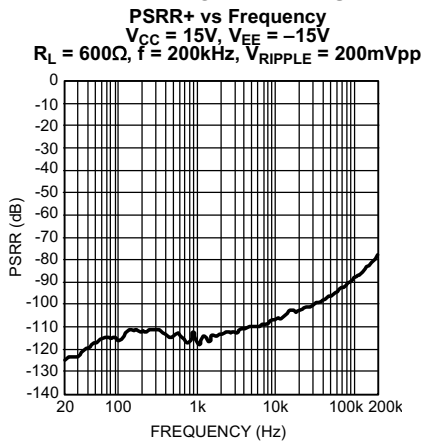


Figure 64.

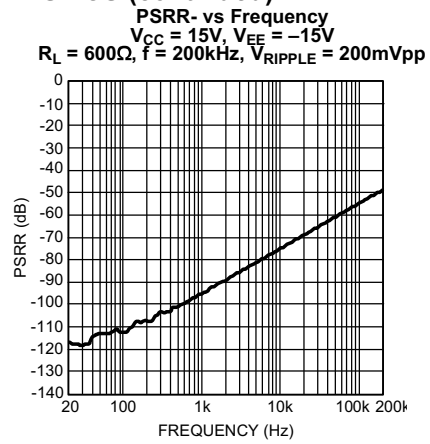


Figure 65.

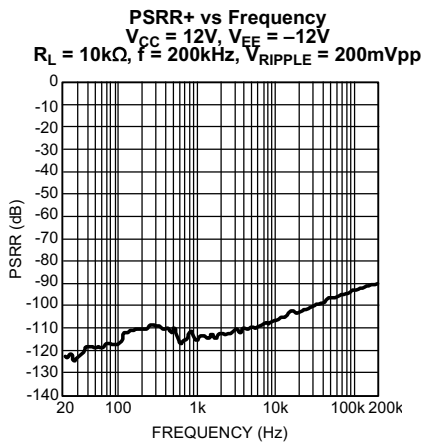


Figure 66.

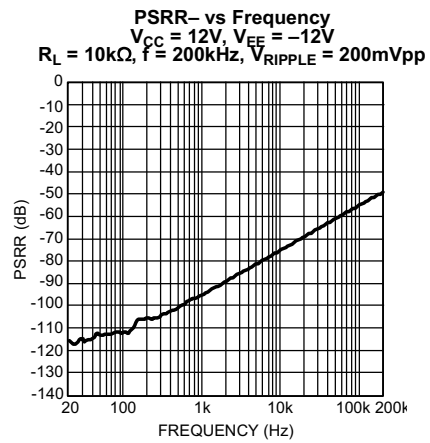


Figure 67.

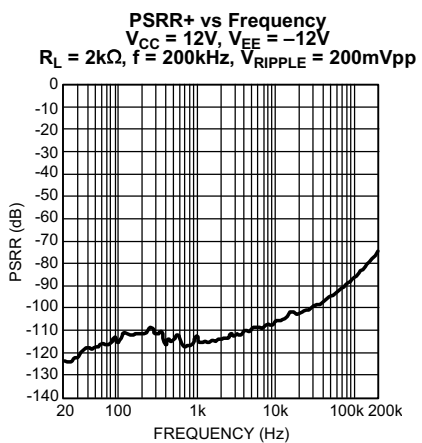


Figure 68.

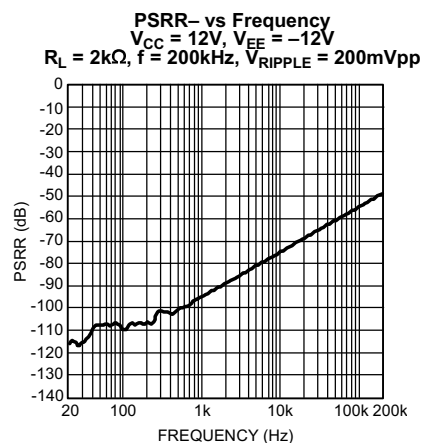


Figure 69.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

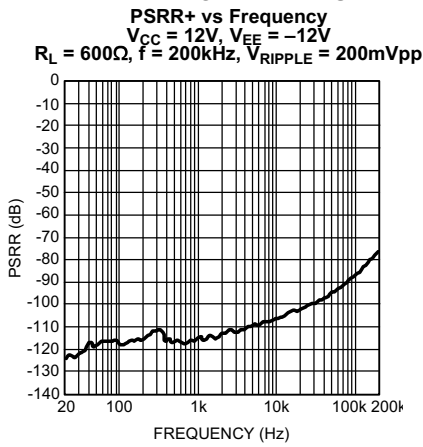


Figure 70.

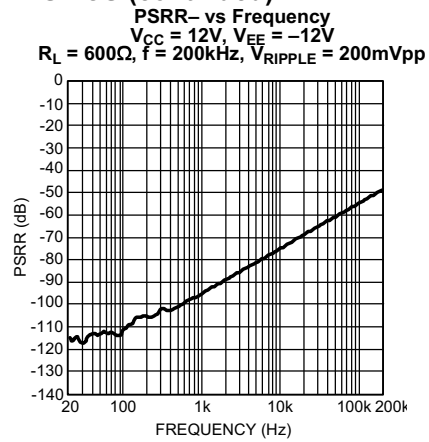


Figure 71.

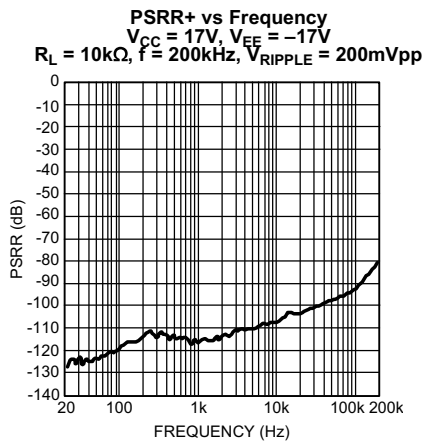


Figure 72.

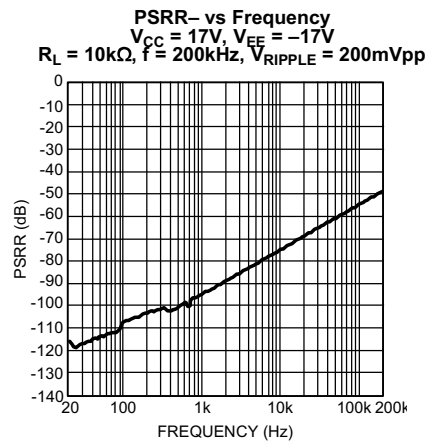


Figure 73.

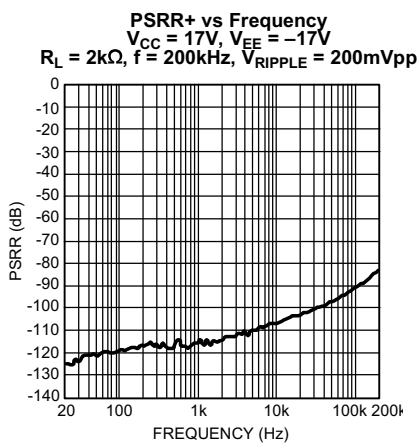


Figure 74.

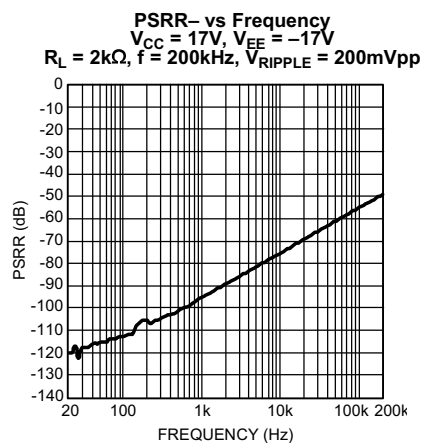


Figure 75.



**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

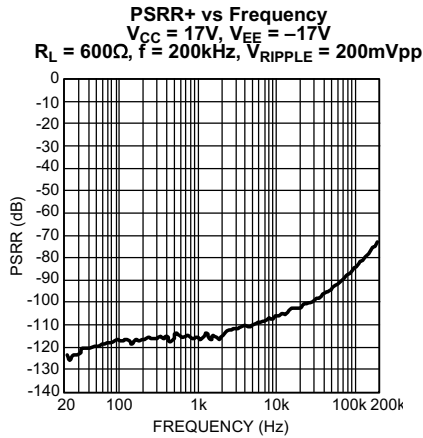


Figure 76.

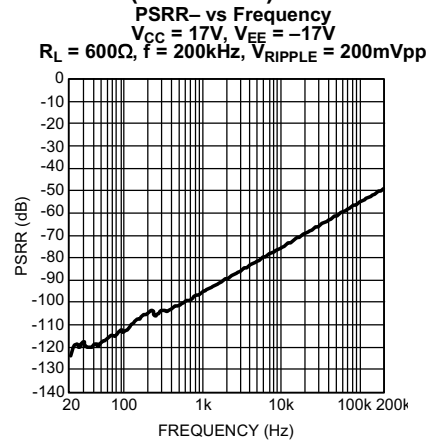


Figure 77.

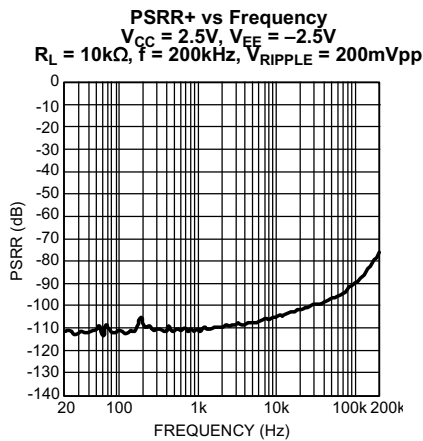


Figure 78.

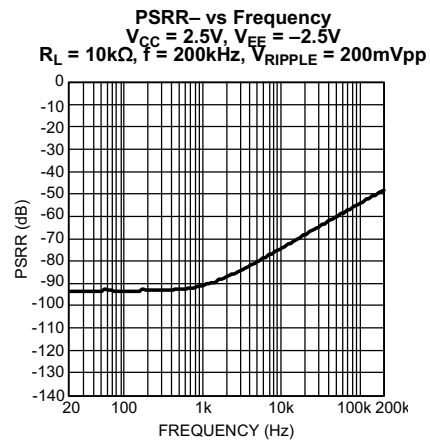


Figure 79.

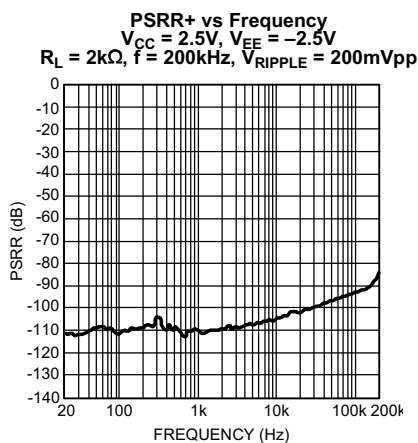


Figure 80.

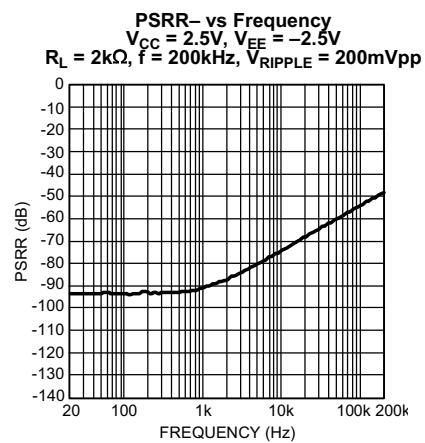


Figure 81.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

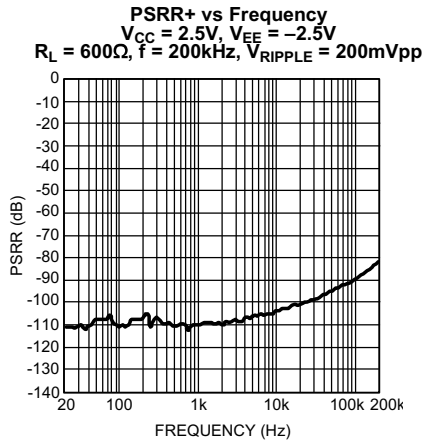


Figure 82.

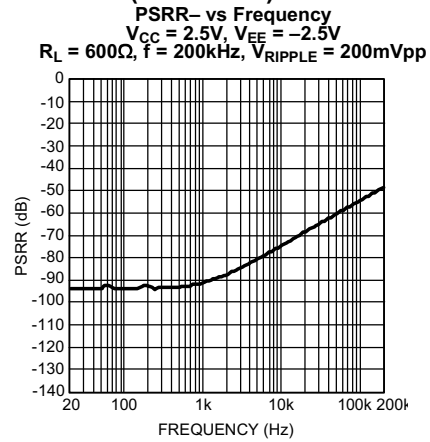


Figure 83.

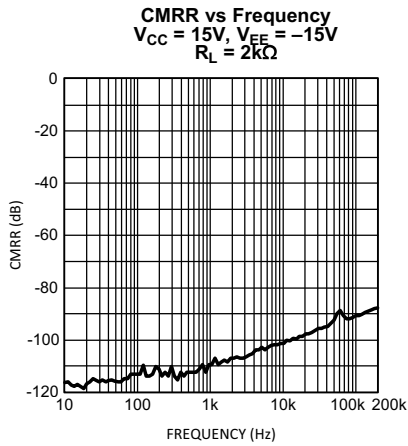


Figure 84.

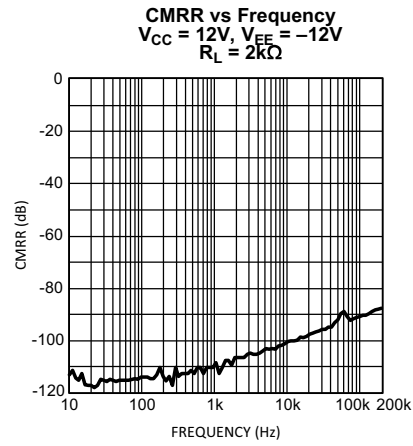


Figure 85.

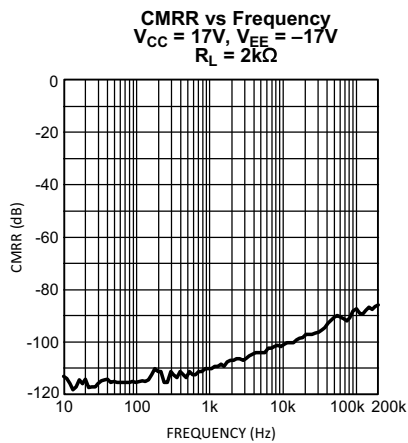


Figure 86.

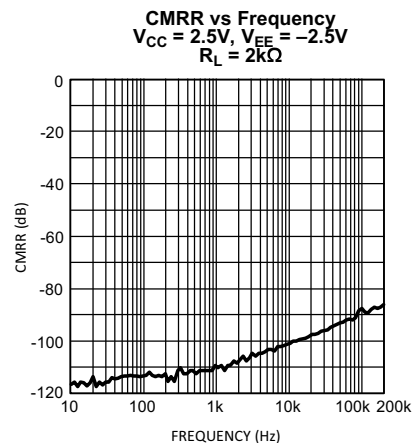


Figure 87.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

**CMRR vs Frequency**  
 $V_{CC} = 15V, V_{EE} = -15V$   
 $R_L = 600\Omega$

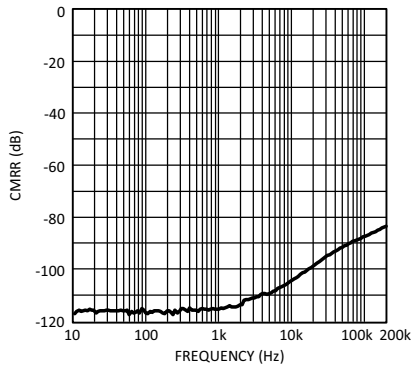


Figure 88.

**CMRR vs Frequency**  
 $V_{CC} = 12V, V_{EE} = -12V$   
 $R_L = 600\Omega$

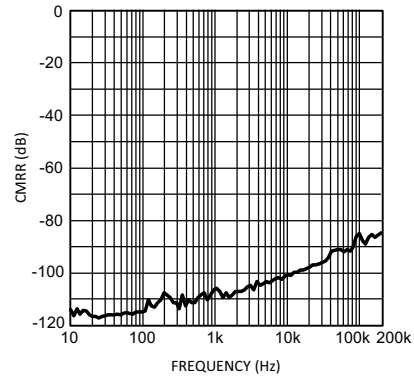


Figure 89.

**CMRR vs Frequency**  
 $V_{CC} = 17V, V_{EE} = -17V$   
 $R_L = 600\Omega$

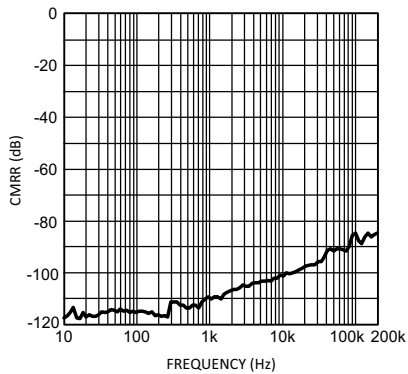


Figure 90.

**CMRR vs Frequency**  
 $V_{CC} = 2.5V, V_{EE} = -2.5V$   
 $R_L = 600\Omega$

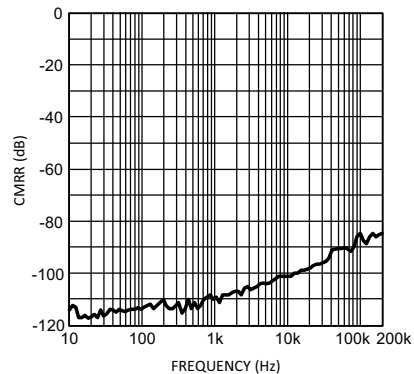


Figure 91.

**CMRR vs Frequency**  
 $V_{CC} = 15V, V_{EE} = -15V$   
 $R_L = 10k\Omega$

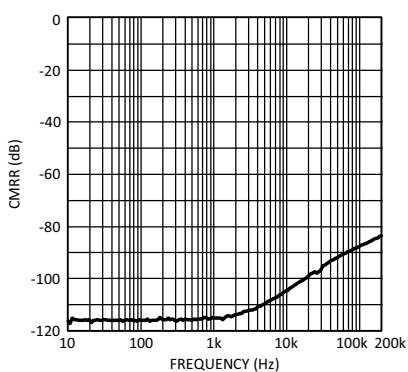


Figure 92.

**CMRR vs Frequency**  
 $V_{CC} = 12V, V_{EE} = -12V$   
 $R_L = 10k\Omega$

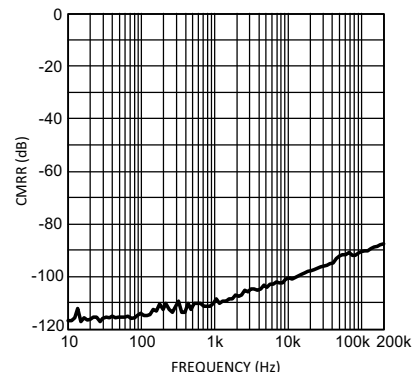


Figure 93.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

**CMRR vs Frequency**  
 $V_{CC} = 17V, V_{EE} = -17V$   
 $R_L = 10k\Omega$

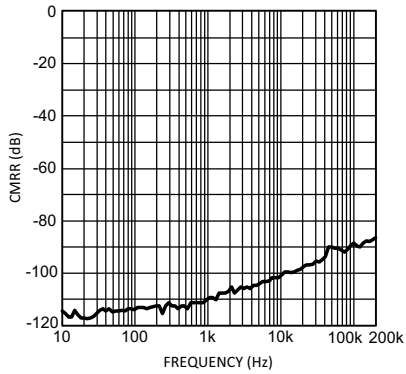


Figure 94.

**CMRR vs Frequency**  
 $V_{CC} = 2.5V, V_{EE} = -2.5V$   
 $R_L = 10k\Omega$

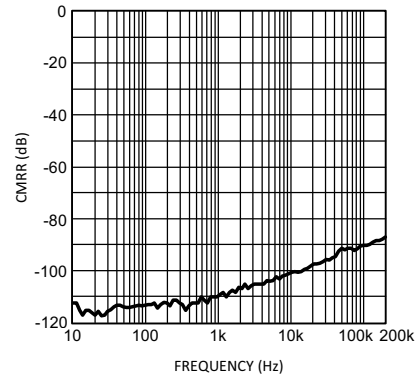


Figure 95.

**Output Voltage vs Load Resistance**  
 $V_{DD} = 15V, V_{EE} = -15V$   
 $THD+N = 1\%$

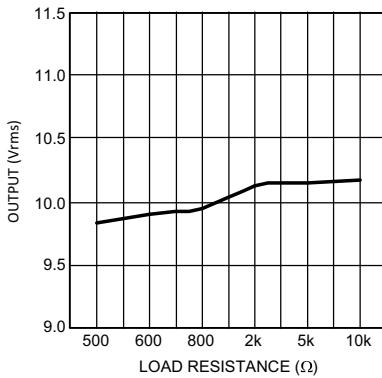


Figure 96.

**Output Voltage vs Load Resistance**  
 $V_{DD} = 12V, V_{EE} = -12V$   
 $THD+N = 1\%$

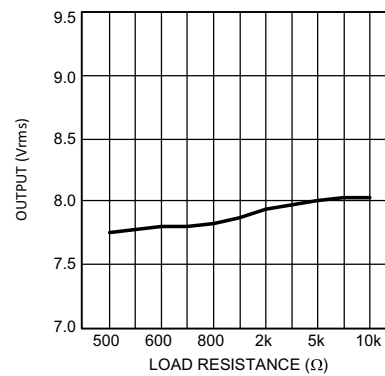


Figure 97.

**Output Voltage vs Load Resistance**  
 $V_{DD} = 17V, V_{EE} = -17V$   
 $THD+N = 1\%$

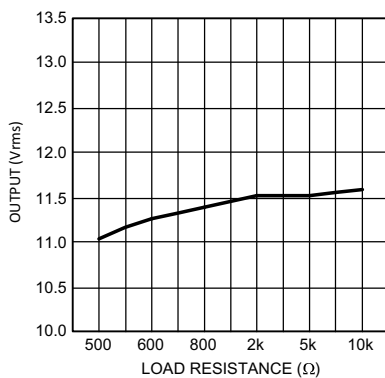


Figure 98.

**Output Voltage vs Load Resistance**  
 $V_{DD} = 2.5V, V_{EE} = -2.5V$   
 $THD+N = 1\%$

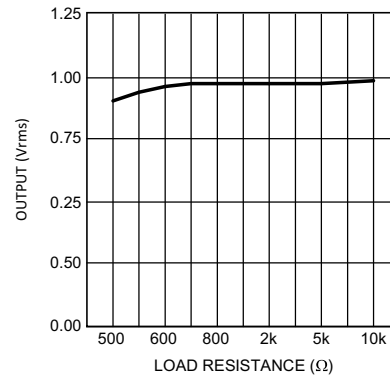


Figure 99.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

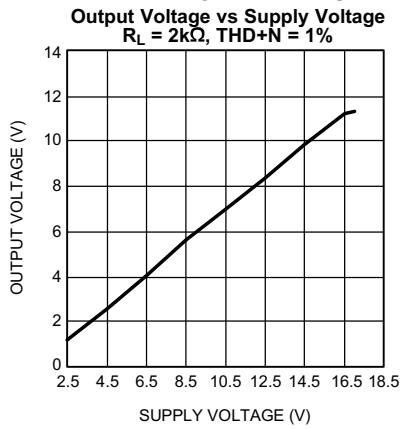


Figure 100.

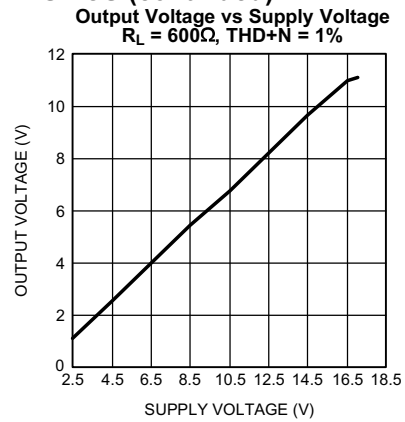


Figure 101.

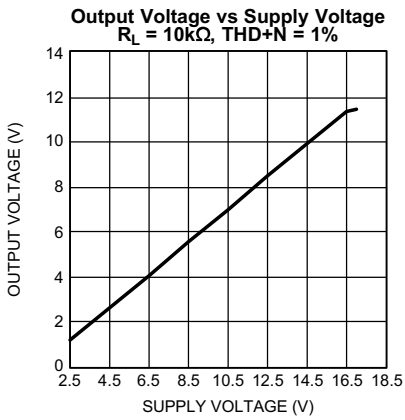


Figure 102.

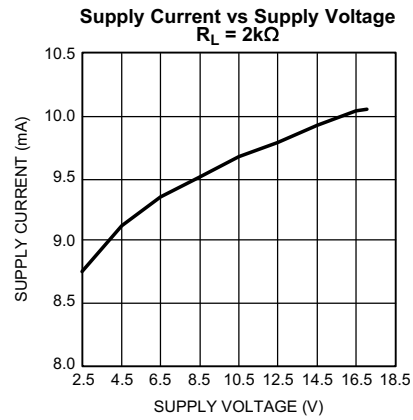


Figure 103.

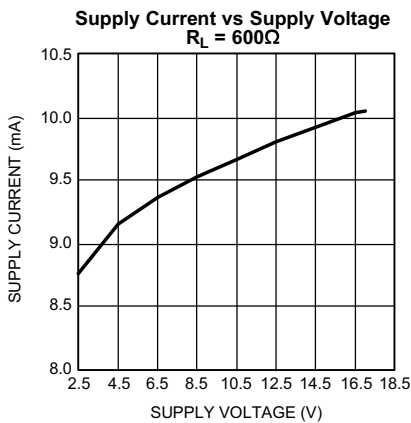


Figure 104.

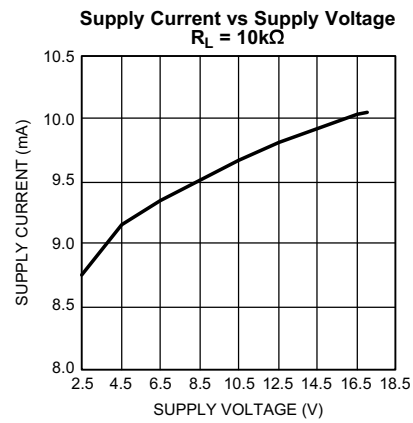


Figure 105.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

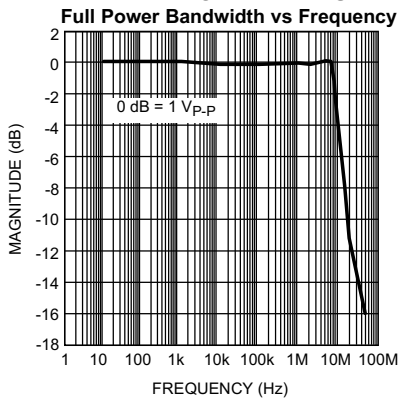


Figure 106.

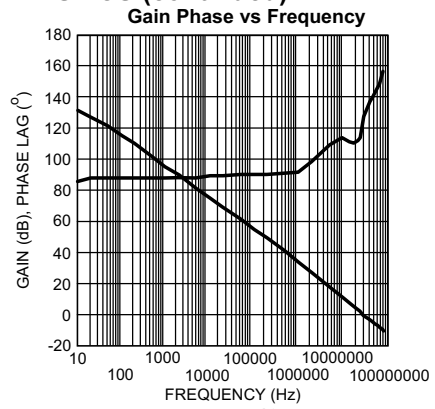


Figure 107.

**Small-Signal Transient Response**  
 $A_V = 1, C_L = 10\text{pF}$

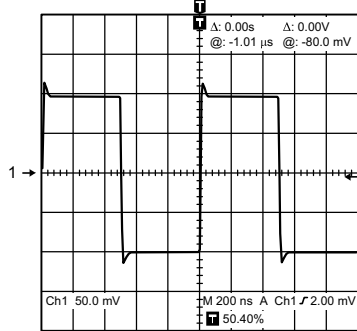


Figure 108.

**Small-Signal Transient Response**  
 $A_V = 1, C_L = 100\text{pF}$

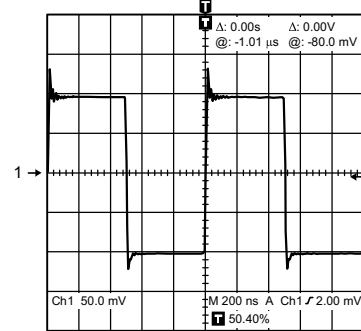


Figure 109.

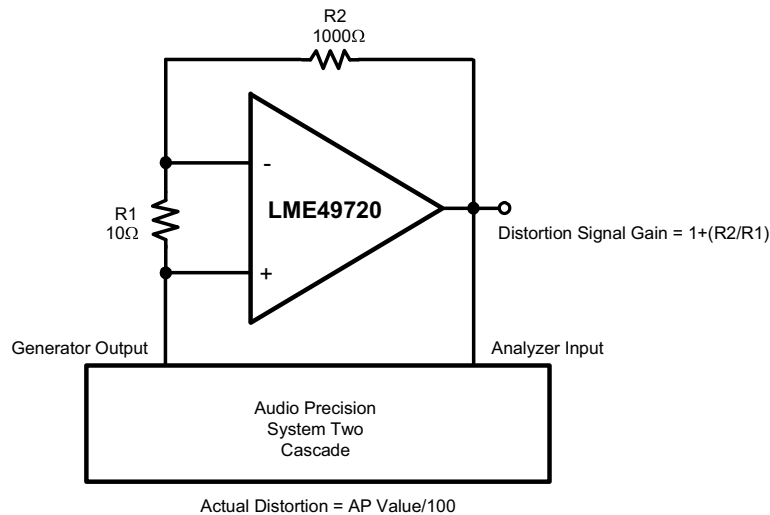
## APPLICATION INFORMATION

### DISTORTION MEASUREMENTS

The vanishingly low residual distortion produced by LME49720 is below the capabilities of all commercially available equipment. This makes distortion measurements just slightly more difficult than simply connecting a distortion meter to the amplifier's inputs and outputs. The solution, however, is quite simple: an additional resistor. Adding this resistor extends the resolution of the distortion measurement equipment.

The LME49720's low residual distortion is an input referred internal error. As shown in Figure 110, adding the 10Ω resistor connected between the amplifier's inverting and non-inverting inputs changes the amplifier's noise gain. The result is that the error signal (distortion) is amplified by a factor of 101. Although the amplifier's closed-loop gain is unaltered, the feedback available to correct distortion errors is reduced by 101, which means that measurement resolution increases by 101. To ensure minimum effects on distortion measurements, keep the value of R1 low as shown in Figure 110.

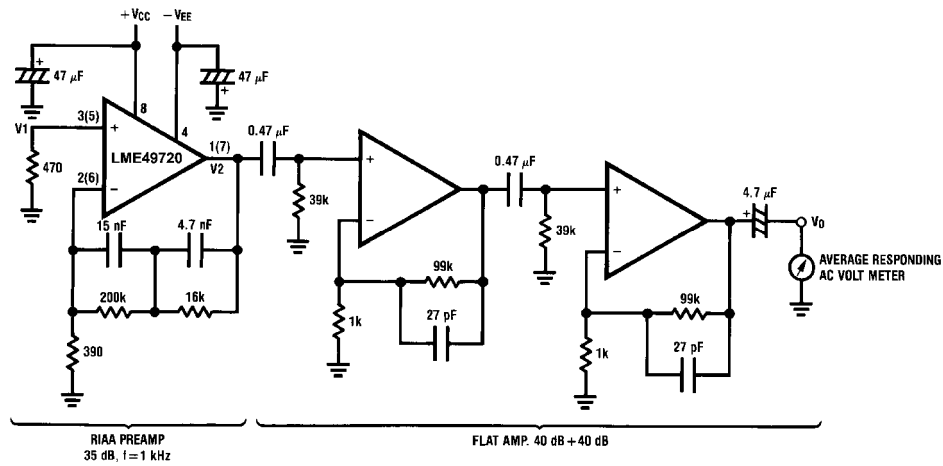
This technique is verified by duplicating the measurements with high closed loop gain and/or making the measurements at high frequencies. Doing so produces distortion components that are within the measurement equipment's capabilities. This datasheet's THD+N and IMD values were generated using the above described circuit connected to an Audio Precision System Two Cascade.



**Figure 110. THD+N and IMD Distortion Test Circuit**

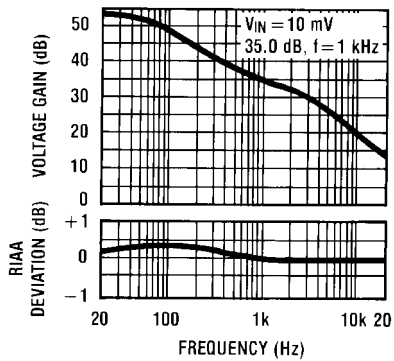
The LME49720 is a high speed op amp with excellent phase margin and stability. Capacitive loads up to 100pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 100pF must be isolated from the output. The most straightforward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.

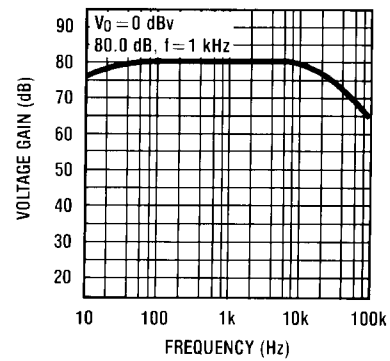


Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.

**Figure 111. Noise Measurement Circuit**  
 Total Gain: 115 dB @f = 1 kHz  
 Input Referred Noise Voltage:  $e_n = V_0/560,000$  (V)



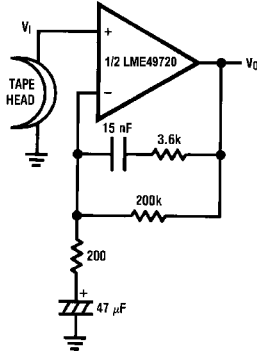
**Figure 112. RIAA Preamp Voltage Gain, RIAA Deviation vs Frequency**



**Figure 113. Flat Amp Voltage Gain vs Frequency**



TYPICAL APPLICATIONS



$A_V = 34.5$   
 $F = 1 \text{ kHz}$   
 $E_n = 0.38 \mu\text{V}$   
 A Weighted

Figure 114. NAB Preamp

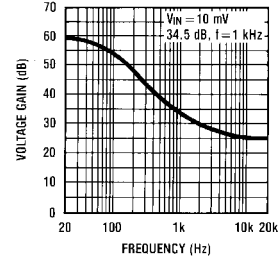
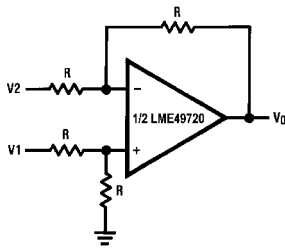
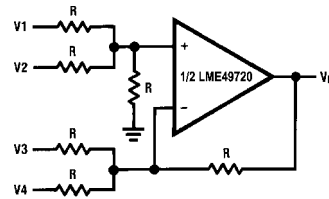


Figure 115. NAB Preamp Voltage Gain vs Frequency



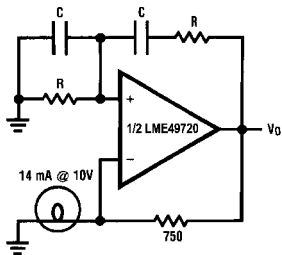
$V_O = V1 - V2$

Figure 116. Balanced to Single Ended Converter



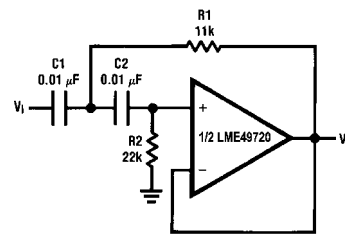
$V_O = V1 + V2 - V3 - V4$

Figure 117. Adder/Subtractor



$f_0 = \frac{1}{2\pi RC}$

Figure 118. Sine Wave Oscillator



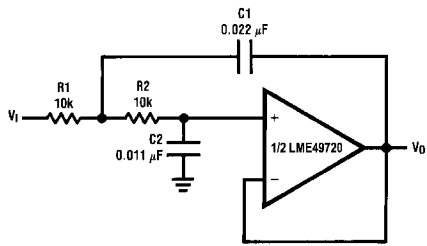
if  $C1 = C2 = C$

$R1 = \frac{\sqrt{2}}{2\omega_0 C}$

$R2 = 2 \cdot R1$

Illustration is  $f_0 = 1 \text{ kHz}$

Figure 119. Second Order High Pass Filter (Butterworth)



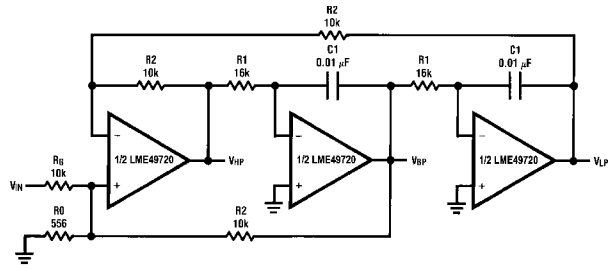
if  $R1 = R2 = R$

$$C1 = \frac{\sqrt{2}}{\omega_0 R}$$

$$C2 = \frac{C1}{2}$$

Illustration is  $f_0 = 1 \text{ kHz}$

Figure 120. Second Order Low Pass Filter (Butterworth)



$$f_0 = \frac{1}{2\pi C1 R1}, Q = \frac{1}{2} \left( 1 + \frac{R2}{R0} + \frac{R2}{R1} \right), A_{BP} = Q A_{LP} = Q A_{LH} = \frac{R2}{R0}$$

Illustration is  $f_0 = 1 \text{ kHz}, Q = 10, A_{BP} = 1$

Figure 121. State Variable Filter

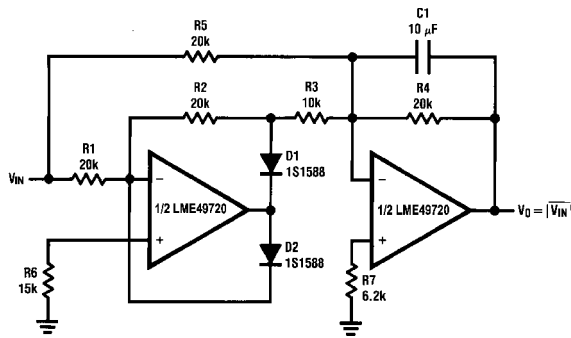


Figure 122. AC/DC Converter

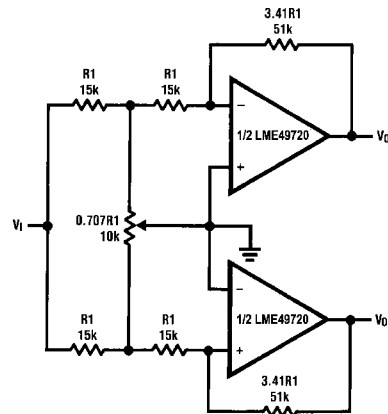


Figure 123. 2 Channel Panning Circuit (Pan Pot)

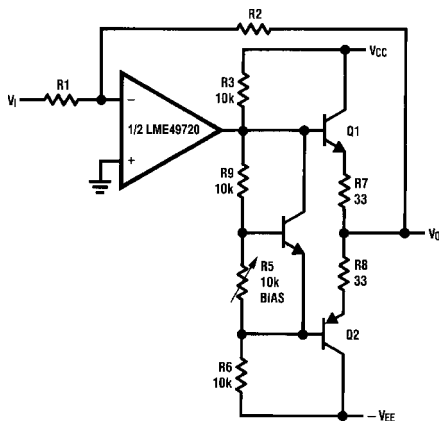
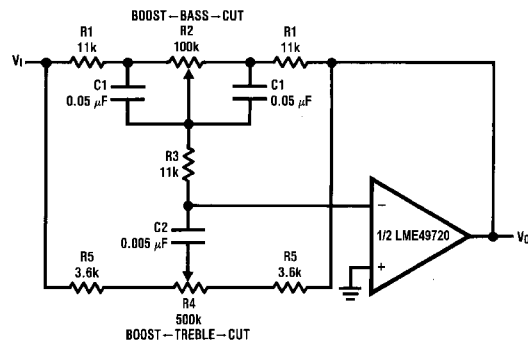


Figure 124. Line Driver



$$f_L = \frac{1}{2\pi R2 C1}, f_{LB} = \frac{1}{2\pi R1 C1}$$

$$f_H = \frac{1}{2\pi R5 C2}, f_{HB} = \frac{1}{2\pi (R1 + R5 + 2R3) C2}$$

Illustration is:

$$f_L = 32 \text{ Hz}, f_{LB} = 320 \text{ Hz}$$

$$f_H = 11 \text{ kHz}, f_{HB} = 1.1 \text{ kHz}$$

Figure 125. Tone Control

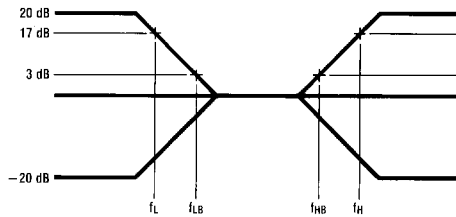
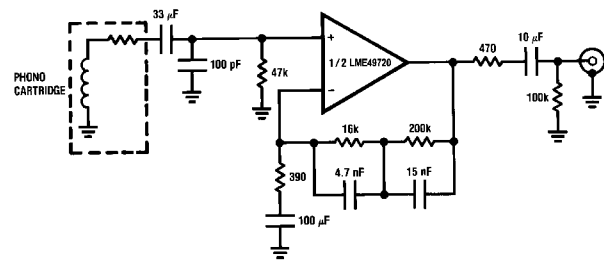
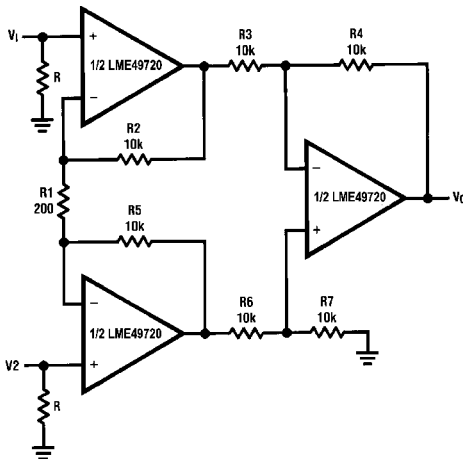


Figure 126.



$A_v = 35 \text{ dB}$   
 $E_n = 0.33 \mu\text{V}$   
 $S/N = 90 \text{ dB}$   
 $f = 1 \text{ kHz}$   
 A Weighted  
 A Weighted,  $V_{IN} = 10 \text{ mV}$   
 @  $f = 1 \text{ kHz}$

Figure 127. RIAA Preamp



If  $R2 = R5, R3 = R6, R4 = R7$   

$$V_0 = \left(1 + \frac{2R2}{R1}\right) \frac{R4}{R3} (V2 - V1)$$
  
 Illustration is:  
 $V_0 = 101(V2 - V1)$

Figure 128. Balanced Input Mic Amp

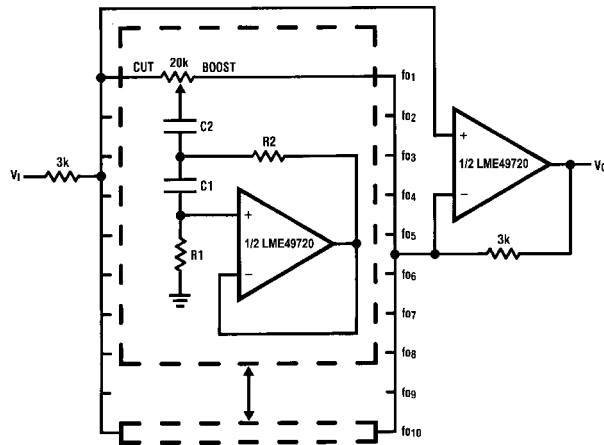


Figure 129. 10 Band Graphic Equalizer

fo (Hz)	C <sub>1</sub>	C <sub>2</sub>	R <sub>1</sub>	R <sub>2</sub>
32	0.12µF	4.7µF	75kΩ	500Ω
64	0.056µF	3.3µF	68kΩ	510Ω
125	0.033µF	1.5µF	62kΩ	510Ω
250	0.015µF	0.82µF	68kΩ	470Ω
500	8200pF	0.39µF	62kΩ	470Ω
1k	3900pF	0.22µF	68kΩ	470Ω
2k	2000pF	0.1µF	68kΩ	470Ω
4k	1100pF	0.056µF	62kΩ	470Ω
8k	510pF	0.022µF	68kΩ	510Ω
16k	330pF	0.012µF	51kΩ	510Ω

**REVISION HISTORY**

<b>Rev</b>	<b>Date</b>	<b>Description</b>
1.0	03/30/07	Initial release.
1.1	05/03/07	Put the "general note" under the EC table.
1.2	10/22/07	Replaced all the PSRR curves.
C	04/05/13	Changed layout of National Data Sheet to TI format.